A U.S. domestic satellite system is on the horizon. But what form will it take? Will it use 4, 6 or 30 GHz? Will it carry telephone calls, television, mail or maybe a magazine-of-the-air? For a review of the various proposals and a look at problems like interference and power limitations, don't miss the report starting on page 36.
Take A New Look
With 3 New CRT’s

NEW 8 x 10 div. CRT—140B main-frame has a full 8 x 10 cm rectangular display area for expanded measurement capabilities.

NEW 8 x 10 div. CRT with variable persistence and storage—in the 141B stops, stores, varies display time in the bigger, rectangular format.

NEW 8 x 10 inch CRT—143A has extra-large display area for group viewing or for monitoring at a distance.

These three new mainframes are in addition to the 140A and 141A mainframes with round CRT’s, and the five mainframes with either round or rectangular CRT’s for spectrum analysis. That makes ten mainframes for better performance in any direction.

Only the HP 140 scope system gives you this combination of capabilities:

Sampling to 12.4 GHz...delayed sweep time base for sampling...50 μV/cm sensitivity dc-coupled with no drift...four channel amplifier...100 μV, 100 dB dual trace...plug-ins for direct readout of TDR, swept-frequency, spectrum analysis—with full compatibility between mainframes and all plug-ins.

With 22 plug-ins, three sampling heads and ten mainframes to choose from, you can match your scope to your measurement requirements. Check the selection chart, see page 508 in your 1969 HP catalog, then contact your HP field engineer. Or, write to Hewlett-Packard, Palo Alto, California 94304. Europe: 1217 Meyrin-Geneva, Switzerland. Price: 140 Mainframes, $640 to $1500; plug-ins $250 to $1900.

<table>
<thead>
<tr>
<th>HP 140 Scope System PLUG-IN CAPABILITIES</th>
<th>VERTICAL PLUG-INS</th>
<th>COMPATIBLE TIME BASES</th>
<th>SPECIAL PLUG-INS</th>
</tr>
</thead>
<tbody>
<tr>
<td>1. Wide Band</td>
<td>140B</td>
<td>1401A</td>
<td>1402A</td>
</tr>
<tr>
<td>2. Sampling</td>
<td>140B</td>
<td>1402A</td>
<td>1403A</td>
</tr>
<tr>
<td>3. High Gain Differential</td>
<td>140B</td>
<td>1402A</td>
<td>1403A</td>
</tr>
<tr>
<td>4. Dual Trace</td>
<td>140B</td>
<td>1402A</td>
<td>1403A</td>
</tr>
<tr>
<td>5. Four Channel</td>
<td>140B</td>
<td>1402A</td>
<td>1403A</td>
</tr>
<tr>
<td>6. X-Y</td>
<td>140B</td>
<td>1402A</td>
<td>1403A</td>
</tr>
<tr>
<td>7. Delayed Sweep</td>
<td>140B</td>
<td>1402A</td>
<td>1403A</td>
</tr>
<tr>
<td>8. No Drift</td>
<td>140B</td>
<td>1402A</td>
<td>1403A</td>
</tr>
<tr>
<td>9. High Common Mode Rejection</td>
<td>140B</td>
<td>1402A</td>
<td>1403A</td>
</tr>
<tr>
<td>10. Algebraic Addition</td>
<td>140B</td>
<td>1402A</td>
<td>1403A</td>
</tr>
<tr>
<td>11. Time Domain Reflectometry</td>
<td>140B</td>
<td>1402A</td>
<td>1403A</td>
</tr>
<tr>
<td>12. Wide Band TDR</td>
<td>140B</td>
<td>1402A</td>
<td>1403A</td>
</tr>
<tr>
<td>13. Swept Frequency</td>
<td>140B</td>
<td>1402A</td>
<td>1403A</td>
</tr>
<tr>
<td>14. Spectrum Analysis</td>
<td>140B</td>
<td>1402A</td>
<td>1403A</td>
</tr>
</tbody>
</table>

HEWLETT PACKARD
OSCILLOSCOPE SYSTEMS
PATENTED PRODUCTS
A "system" for testing tantalum and aluminum electrolytics

The Type 1617 Capacitance Bridge is really a system in one package designed specifically to test electrolytic capacitors at 120 Hz per MIL or EIA specifications. However, this 1% bridge can measure any capacitor, including those as large as 1.1 F. Besides having an exceptionally wide C range, the 1617 has many other features that make it an excellent general-purpose bridge for component testing at quality-control and incoming-inspection stations.

WIDE RANGE. C range of $10^2$, from 1 pF to 1.1 F, with an accuracy of $\pm 1\% \pm 1$ pF and 2% from 0.11 F to 1.1 F. D range from 0 to 10 with an accuracy of approximately $\pm 2\% \pm 0.001$.

CONVENIENCE. The 1617 is completely self-contained; just plug it into a power line and start measuring. It includes a 120-Hz generator, a tuned detector, and an adjustable dc polarizing voltage, all enclosed in a handy, portable, flip-tilt carrying case.

MIL and EIA SPECIFICATIONS. It meets or exceeds the requirements of: MIL-C-39003 (Solid Tantalum), MIL-C-3965C (Tantalum Foil and Sintered Slug), MIL-C-39018 (Aluminum Oxide), (EIA) RS 154B (Dry Aluminum), MIL-C-62C (Polarized Aluminum), (EIA) RS 205 (Electrolytic), MIL-C-26655B (Solid Tantalum), (EIA) RS 228 (Tantalum).

OTHER FEATURES. The 1617 bridge:
- has a metered, dc polarizing voltage that is adjustable from 0 to 600 volts; external bias up to 800 volts may be applied.
- provides for 2-, 3-, 4-, or 5-terminal connections to minimize the effects of residual impedances.
- detects leakage current down to 0.5 µA.
- can operate at frequencies up to 1000 Hz with an external generator.
- has important safety features, such as warning lights and discharge circuitry, which protect both the operator and the instrument.
- has Orthonull® balance finder to permit rapid bridge balances and eliminate sliding nulls during high-D capacitance measurements.
- has a generator that is phase-reversible to reduce the effects of hum pickup; amplitude is selectable and limited to 0.2 V, 0.5 V, or 2 V.

The 1617 bridge can also measure the capacitance and loss of cables, transformers, insulating materials, and electric motors.

Price: $1250 in the U.S.A. For complete information, write General Radio, W. Concord, Massachusetts 01781; telephone (617) 369-4400. In Europe: Postfach 124, CH 8034 Zurich 34, Switzerland.
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Two of 145
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Pulse generators  Spectrum analyzers
Microwave frequency  Digital panel meters
Indicators  Microwave signal
Digital clocks  generators
Memory testers  Laboratory magnets
Analog computers  Data acquisition
Time code generators  systems
Data generators  Microwave test sets

Input filtering PLUS dual-slope integration form a unique double barrier against noise in these fast-reading DVM's. This new idea from S-D gives you more accuracy for less money than any other .01% DVM.

The portable Model 9200 and the Thin-Line, rack-mounting Model 9300 come with all the popular ranges, functions and interfacing options. Prices start at $1175.

Send for the complete, proven specs and see for yourself. Contact Measurements Division, Systron-Donner Corporation, 888 Galindo Street, Concord, Calif. 94520. Phone (415) 682-6161.

Newest DVM idea: double noise barrier boosts accuracy!
NEWS
21 News Scope
25 From 8-1/3 bps to 16,200 bps—Here’s how
New telemetry for Mariner began as experiment and ended with the
best pictures yet taken of Mars
30 Computerized ‘eye’ to shed light on the sun
16 pre-programed investigations will be available when Air Force puts
new solar telescope to work
32 Animal tests to help evaluate TV radiation
36 What’s delaying U.S. satellite communications?
A special report
51 Washington Report
56 Sidelights
59 Editorial: Launch—or get off the pad

TECHNOLOGY
68 Develop useful general models of nonlinear devices for computer aided
design. The tunnel diode illustrates the curve-fitting approach.
74 Streamline feedback-amplifier design. It’s simple if you analyze forward and
reverse transmission paths of your circuit separately.
78 One step ahead of need (never two) was the over-all managing technique used
by this Grumman engineer who directed NASA’s assembling and testing program
for LM 5.
85 Ideas for Design

PRODUCTS
94 Instrumentation: Two new instruments display with LED numerics.
100 Components: Tiny balanced mixer spans 25 kHz to 500 MHz.
97 ICs & Semiconductors
103 Data Processing
104 Modules & Subassemblies
106 Microwaves & Lasers
107 Packaging & Materials
108 Production

Departments
14 Designer’s Datebook
109 Evaluation Samples
110 Design Aids
110 Annual Reports
111 Application Notes
112 New Literature
118 Advertisers’ Index
120 Information Retrieval Service

Cover Photo: Courtesy of TRW’s Systems Group, Redondo Beach, Calif.
The Intelsat III satellite has a mechanically despun antenna.
Put Your New Silicon Power Circuit Designs In The Best Shape Possible Now!

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High reliability, high frequency and high current for NPN or PNP switch/amplifier designs to 300 V

Shape 3: EpiBase Silicon Power
Efficient, low-cost key to replacement of high-current germanium in PNP and NPN to 140 V

Shape 5: Epitaxial Annular Silicon Power
Double-diffused for high-current reliability and up to 1,000 times less leakage

Shape 6: Hybrid Single-Diffused Silicon Power
No. 1 industrial choice for efficient, lower-cost, rugged switching/amplifying to 140 V

MOTOROLA
Silicon Power Transistors

INFORMATION RETRIEVAL NUMBER 4

ELECTRONIC DESIGN 20, September 27, 1969

*Trademark Motorola Inc.
†Annular Semiconductors Patented by Motorola Inc.
The Trend is TTL.
Icon all-digital numerical control helps customer cut costs and downtime.

TI integrated circuits help Icon build reliability and economy into all-digital numerical controls.

Icon Corporations' System 350 all-digital open-loop numerical control is turning in new records for reliability, economy and ease of installation.

The TI Series 74N TTL integrated circuits used in the system are giving flawless performance. About the ICs, Dr. Gordon Baty, Icon president said: "Of the more than 30,000 we have in the field, we've had almost no inexplicable failures. We think this kind of trouble-free performance is what machine tool users want. We sink or swim on reliability and performance on the machine shop floor."

Icon's open-loop system offers far greater reliability, lower overall costs, and easier retrofit of existing machine tools.

One retrofit example: National Connector purchased a 75-year old vertical mill from a junkyard for $250. They restored it and installed an Icon System 350 numerical controller.

Total costs ran about $15,000 (less than one-half the cost of machines with similar capabilities). The restored mill met both tolerance and production-rate demands of National's high-precision interconnection manufacturing facility.

Retrofit went smoothly and quickly. Both mill and controller worked perfectly the first time with normal adjustments. Today, after more than a year, the system is still working perfectly.

Maintenance is simplified. High-density IC design packs instantly recognizable functions onto individual cards—allowing fast diagnosis and easy replacement.

Each Icon System 350 utilizes more than 250 TI Series 74N TTL integrated circuits in low-cost plastic packages. About twenty different types are used—including some of the latest MSI circuits.

New design aid

Integrated circuits from TI are helping many other manufacturers of industrial systems to increase reliability, improve performance, and reduce costs. Find out more from a new 80-page brochure that includes valuable design information on Series 54/74—industry's broadest TTL line. Circle 118 or write Texas Instruments Incorporated, P.O. Box 5012, M.S.308, Dallas, Texas 75222.
Indicator tubes—simple interfacing, limited flexibility, low in cost.

Full-scale terminals—extremely flexible, very expensive.

One way Two way
New kind of CRT Readout—the best of two worlds.

New way

Announcing the SRD-100—a new CRT readout device with simple interfacing. Now it's practical to use CRT readout displays in place of indicator tubes. Without getting into the cost of full-scale terminals. Especially if you need remote or multiple displays. And even if you need only 5-10 digits.

The SRD-100 accepts numeric/symbolic BCD data in parallel form, converts it to a composite video signal and displays it on any 525-line TV monitor. In any of 160 character positions. Interfacing is as simple as indicator tubes.

And you can display data at remote locations with any number of CRTs—with just a single coaxial cable.

You can blink data off and on four times per second. Use three different sizes of character display. And reverse the character and display background from black on white to white on black. All controlled independently, line by line.

Cost for the basic SRD-100, not including TV monitors, is $995. Characters in groups of half lines, plus options, are additional.

So if you have a numerical readout problem in any application like automatic testing, process control, automation or data acquisition, you're no longer limited to indicator tubes. Now you have a choice.

From the company that specializes in innovative data communications systems.

Mr. Richard Kaufman
Marketing Director, Dept. 11
Applied Digital Data Systems, Inc.
89 Marcus Boulevard
Hauppauge, New York 11787

Please send more information about the SRD-100.

Name________________________
Title________________________
Company_____________________
Address_____________________
City_________________________
State________________________
Zip__________________________
Mini-est mfd for μ spaces
or: our next series may be invisible

You can still see our new Minitan® W-Series tantalum capacitors... but just barely. These extraordinary little solid-electrolyte devices — the industry's smallest — pack up to .47 mfd. into a case about the size of a pin head.

What's more, they do it with a maximum DC leakage (at 25°C) of only 0.5μA, standard tolerances to ±5% and a 130% surge voltage rating. Gold-plated solid nickel leads and an operating temperature range of -55°C to +85°C help make this the finest series of microminiature modular capacitors available for hybrid and thick film circuit use.

Considerably smaller than comparably rated CS13 and epoxy filled devices, Minitan® W's have also out-shrunk monolithic ceramics. For example, a typical .22 mfd. ceramic measures .350 x .095 x .070; the Minitan® W case is only .100 x .050 x .040.

Of course, if you like capacitors you can see with scarcely a second glance, Components also has the broadest line of CS13, CSR13, CSR09, subminiature, and microminiature modular, cordwood, and non-polars around. Send for our new general catalog and get the small picture.

---

**W CASE SIZE**

<table>
<thead>
<tr>
<th>PART NUMBER</th>
<th>CAP. IN MFD @ 25°C 120 Hz</th>
<th>WVDC @ 85°C</th>
</tr>
</thead>
<tbody>
<tr>
<td>AXIAL</td>
<td>RADIAL</td>
<td></td>
</tr>
<tr>
<td>W472A</td>
<td>W472R</td>
<td>.0047</td>
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<tr>
<td>W682A</td>
<td>W682R</td>
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<td>.010</td>
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<td>W153A</td>
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<td>.015</td>
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<td>W223A</td>
<td>W223R</td>
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<td>W473A</td>
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<tr>
<td>W104A</td>
<td>W104R</td>
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<tr>
<td>W154A</td>
<td>W154R</td>
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<tr>
<td>W224A</td>
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<td>.22</td>
</tr>
<tr>
<td>W334A</td>
<td>W334R</td>
<td>.33</td>
</tr>
<tr>
<td>W474A</td>
<td>W474R</td>
<td>.47</td>
</tr>
</tbody>
</table>

**DIMENSIONS**

INCHES

- A: .100 max.
- B: .050 max.
- C: .040 max.
- D: .030 ±.016

**MINITAN® MODULAR**

(Also available with axial leads)

**MINITAN® CORDWOOD**

---

**COMPONENTS, INC.**

BIDDEFORD, MAINE 04005 - TEL. (207) 284-5956 - TWA 710-229-1559

INFORMATION RETRIEVAL NUMBER 6

INFORMATION RETRIEVAL NUMBER 7
Pick your own happy ending.

All good things, including circuit boards, must come to an end. That's where we come in. We've got the happy ending patented. Four times. (Pat. nos. 2,248,686; 2,294,056; 3,182,276; 3,208,026.) In all the configurations you need. Four plug styles. Ten receptacle styles. For direct staking, or with their own insulator. With or without card guides. With or without guide pins. With any number of contacts (up to 135, on one triple-decker receptacle).

Every one meets Mil-E-5400 including the new revision “K”, which makes metal-to-metal contacts mandatory.

Our Varicon™ metal-to-metal contacts are patented and famous, in their quiet way. Because, in millions of operating hours, not one has failed. The Varicon contacts have a unique fork-like design, with four large, coined, mating surfaces that form a firm gas-tight seal. Even after being mated for years, they remain clean and unoxidized. They have, and keep, a low contact resistance and high current capacity. And they're exceptionally resistant to shock and vibration.

Several members of our p.c. connector family come with our Variloc™ contact. These have all the advantages of the Varicon contacts — except that you can insert and remove them from the receptacle yourself. They're available by the handful for small scale production and on 1800-contact reels for use with automatic crimping equipment. They're all described in our printed circuit connector guide. To get your copy, write, wire, call or TWX us. Elco Corporation, Willow Grove, Pa. 19090. (215) 659-7000. TWX 510-665-5573.
Gates
DM8000N (SN7400N) Quad 2-Input, NAND gate
DM8001N (SN7401N) Quad 2-Input, NAND gate (Open Collector)
DM8002N (SN7402N) Quad 2-Input, NOR gate
DM8003N (SN7403N) Quad 2-Input, NAND gate (Open Collector)
DM8004N (SN7404N) Hex inverter
DM8010N (SN7410N) Triple 3-Input, NAND gate
DM8020N (SN7420N) Dual 4-Input, NAND gate
DM8030N (SN7430N) Eight-Input, NAND gate
DM8040N (SN7440N) Dual 4-Input, Buffer
DM8050N (SN7450N) Expandable Dual 2-Wide, 2-Input AND-OR-INVERT gate
DM8051N (SN7451N) Dual 2-Wide, 2-Input AND-OR-INVERT gate
DM8053N (SN7453N) Expandable 4-Wide, 2-Input AND-OR-INVERT gate
DM8054N (SN7454N) Four-Wide, 2-Input AND-OR-INVERT gate
DM8060N (SN7460N) Dual 4-Input expander
DM8086N (SN7486N) Quad Exclusive-OR-gate

Flip Flops
DM8540N (SN7472N) MASTER-SLAVE J-K flip flop
DM8501N (SN7473N) Dual J-K MASTER-SLAVE flip flop
DM8500N (SN7476N) Dual J-K MASTER-SLAVE flip flop
DM8510N (SN7474N) Dual D flip flop

Counters
DM8530N (SN7490N) Decade Counter
DM8532N (SN7492N) Divide-by-twelve counter
DM8533N (SN7493N) Four-bit binary counter
DM8560N (SN74192N) Up-down decade counter
DM8563N (SN74193N) Up-down binary counter

Decoders
DM8840N (SN7441AN) BCD to decimal nixie driver
DM8842N (SN7442N) BCD to decimal decoder

Miscellaneous
DM8550N (SN7475N) Quad latch

Spec-for-spec. Pin-for-pin. Second source on all popular 74N.
These exclusive National TTL MSI circuits offer designers increased opportunities for circuit savings in design time, board space and costs. Write or call for your free TTL Spec Guide, TTL Cross Reference Guide, and TTL Performance Tables. National Semiconductor, 2975 San Ysidro Way, Santa Clara, California 95051 (408) 245-4320 TWX: 910-339-9240 Cables: NATSEMICON
There are obvious advantages in speed, cost and often in efficiency of
the ultimate system in having one source produce both miniature lamps
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We are major suppliers of fiber optics and incandescent illumination
sources to principal computer manufacturers and other light guide
users.

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optic and photo-sensitive device applications with selectable lens-
filament combinations and basing for optimum utilization of energy.

Our lamp-fiber optic systems are designed for maximum efficiency
and are often less costly than standard multiple-lamp systems.

Phone or write for more information

WELCH ALLYN, INC.
Skaneateles Falls, N.Y. 13153
(315) 685-5788
When your 1969 model looks the same as your 1968 model—getting across the engineering advances and improvements that are on the inside is a problem.

For instance, in 1966 we started using fully-annealed Armco steel for all Guardian Solenoid plungers... an "inside" improvement. Then, to compound the problem, we covered up this improvement with copper/nickel plating.

In 1968 we did it again. We took those long-life plungers and started running them in a cavity lined with low-friction phenolic. This alone increases operating life by maybe half a million operations.

And there's more: The new acetate-yarn-sealed coil cover that's standard this year means better protection, complies with U/L construction at no extra cost.

Our "bug" changes. Inside. Where an engineering advance makes for a better solenoid. Write for Bulletin G2, TS.
MC1595 is the industry's first true Linear, 4-Quadrant Multiplier IC.

Here is the first element in what promises to be a large family of monolithic, linear multipliers. The MC1595 is designed for uses where the output voltage is a linear product of two input voltages; and, as such, its list of applications is almost limitless, particularly in the control and instrumentation fields.

For example, some of the applications are:

- To Multiply
- To Divide
- To Find Square Root
- To Determine Mean-Square

- Frequency Doubling

The MC1595 can also be used as a Balanced Modulator/Demodulator, and for Electronic Gain Control among many others. It even has the capability for determining true rms; plus direct power calculations.

Here are some of the features that contribute to the wide versatility of the MC1595:

- Excellent Linearity — 1% max error “X” input; 2% error max “Y” input.

- Wide Bandwidth — Phase Error \( \leq 3^\circ \) from DC to 750 kHz.
- Adjustable Scale Factor.
- Large Input Voltage Range \( \pm 10 \) V.

The MC1595 Multiplier is currently available from distributor stock in the 14-pin dual in-line ceramic package. So is its 0 to 70° temperature-range counterpart, the MC-1495. For complete information, including a just-published Application Note, circle the reader service number, or write: P. O. Box 20912, Phoenix, Arizona 85036.
NEW 7000 SERIES
Plug-In Oscilloscopes

150 MHz Bandwidth

USABLE performance to 150 MHz or 90 MHz. Combined mainframe and plug-in bandwidths are specified at minimum deflection factors with or without probes. With . . .

MORE Sensitivity

Higher sensitivities are achieved at greater bandwidths than ever before. 5 mV/div at 150 MHz, 1 mV/div at 100 MHz and 10 µV/div at 1 MHz. With . . .

MORE Flexibility

Each mainframe accepts up to four plug-in units. Thirteen plug-ins are currently available to cover virtually all multi-trace, differential, sampling, and X-Y applications. Plus . . .

NEW Convenience

Greater convenience in all areas of instrument operation. Features such as Auto Scale Factor Readout, lighted push-button switching, and true automatic triggering assure faster, more accurate, less complicated measurements.

Please turn for additional information.
CONTROLS
At "A" sweep, "B" sweep, independently. A single-astigmatism adjustment, under complete the control

CALIBRATOR
A multi-function generator usable as a "standard" for calibration of voltage and current GAIN, time/div, and probe compensation. The output is DC or AC (1 kHz or variable) voltage or current (fixed at 40 mA). The amplitude accuracy is within 1% and the time accuracy is within 0.5% at 1 kHz.

TRIGGERING
The signals from both vertical plug-ins are coupled through a mainframe logic circuit and made available to each horizontal plug-in, selectable from LEFT channel, RIGHT channel, or slaved to VERTICAL MODE. The latter frees the operator from manual source changes during single-trace operation and, in conjunction with the P-P AUTO TRIGGER MODE in the time-base units, provides true hands-off triggering during routine measurements.

FOUR PLUG-IN CHANNELS
The modular approach is the answer to instrument flexibility. With dual-trace switching in the mainframe amplifiers, each plug-in can be "specialized" in function and operate in combination with other units. Thirteen plug-ins are currently available for the 7000-Series. Together, they represent the widest range of performance options for multi-trace, differential and sampling applications available today.

7A22 High-Gain Differential Amplifier
Bandwidth—DC to 1 MHz with selectable upper and lower —3 dB points.
Min deflection factor—10 µV/div at full bandwidth.

7M11 Delay Line Unit
Two 75 ns, 50-Ω delay lines. Trigger selection from either line.

7B51/7B50 Time-Base Units for the 7504
5 ns/div maximum sweep speed. Operable singly or in combination for delaying sweep capability.

7S11 Sampling Amplifier
Accepts the plug-in sampling heads for bandwidths to 14 GHz (25 ps tr).

7T11 Random Sampling Time Base
10 ps/div to 5 ns/div sweep range, accomplished with equivalent-time and real-time techniques. Triggering to 12 GHz.
**AUTO SCALE FACTOR READOUT**

A character generator senses the position of volts/div, amps/div, time/div, polarity, and uncalibrated variable controls, then accounts for probe attenuation and displays the correct scale factors for all channels directly on the CRT.

**BRIGHT TRACE**

The acceleration potentials are 24 kV for the 7704 and 18 kV for the 7504 for improved trace visibility. Single-shot photographic writing speed is 3300 cm/µs (7704) measured with the standard P31 phosphor, the new C-51 camera and 10,000 ASA film. The display area is 8 cm x 10 cm with a parallax-free illuminated graticule.

**DUAL-TRACE SWITCHING**

Both the vertical and horizontal mainframe amplifiers are "dual trace" providing a unique level of flexibility with plug-in combinations. A relatively small number of plug-ins can then meet a wide range of application requirements. The CHOP and ALT modes permit simultaneous displays of delaying and delayed sweep, and, through switching logic, may be "slaved" to provide a functional dual-beam type of display.

**7A13 Differential Comparator Amplifier**

Bandwidth—DC to 100 MHz (3.5 ns tr) in the 7704; DC to 75 MHz (4.7 ns tr) in the 7504.
Min deflection factor—1 mV/div at full bandwidth.

**7B71/7B70 Time-Base Units for the 7704**

2 ns/div maximum sweep speed. Operable singly or in combination for delaying-sweep capability.

**7A11 Captive FET Probe Amplifier**

Bandwidth—DC to 150 MHz (2.4 ns tr) in the 7704; DC to 90 MHz (3.9 ns tr) in the 7504.
Min deflection factor—5 mV/div at full bandwidth.

**7A12 Dual-Channel Amplifier**

Bandwidth—DC to 105 MHz (3.4 ns tr) in the 7704; DC to 75 MHz (4.7 ns tr) in the 7504.
Min deflection factor—5 mV/div at full bandwidth.

**7A14 AC Current Probe Amplifier**

Bandwidth—25 Hz to 105 MHz depending on mainframe and current probe; two probes available.
Min deflection factor—1 mA/div at full bandwidth.
C-51/C-50 Trace-Recording Cameras

Two new compact trace-recording cameras have been designed for direct compatibility with the 7000-Series Oscilloscopes. The C-51 and C-50 cameras are basically identical units, differing only in the lens system. The C-51 has an f/1.2, 1:0.5 lens; the C-50 uses an f/1.9, 1:0.7 lens. The C-51 is recommended for single-shot photography at the fastest sweep rates, the C-50 for more general purpose applications. Photographic writing speed of the two 7000-Series mainframes with the C-51 and 10,000 ASA film (without prefogging) is 3300 cm/µs (7704) and 2500 cm/µs (7504).

The cameras offer a new level of operational convenience for mistake-proof trace photography. The guess work normally associated with selection of f stop and shutter speed to match the ASA index and trace brightness is eliminated. After setting the ASA index, the built-in photometer allows a visual correlation of trace intensity to the correct f stop setting and shutter speed. After initial adjustment, a change of f stop or shutter speed will still maintain the same exposure. Focusing is accomplished by two beams of light projected on the CRT which, when superimposed, indicates optimum focus. The insert shows the photometer spot and the rangefinder focusing images.

SCOPE-MOBILE® CARTS
The 204-2 Scope-Mobile® Cart is specifically designed for the 7000-Series instruments. It provides a securing mechanism for the oscilloscope, nine positions of selectable tray tilt, a large storage drawer, storage for five 7000-Series plug-ins, and large locking-type wheels.

PROBES
The P6053 is a miniature fast-rise 10X probe designed for full compatibility with the 7000-Series instruments. Input R and C is 10 MΩ, 10.3 pF. Probe risetime is 1.2 ns or less.

The P6052 is a passive dual-attenuation probe designed for measurements below 30 MHz. A sliding collar selects 1X or 10X attenuation. Input R and C is 1 MΩ or 10 MΩ, 100 pF or 13 pF. Risetimes are 60 ns (1X) and 7 ns (10X).

Tektronix, Inc.

For a demonstration, call your local Tektronix field engineer or write: Tektronix, Inc., P.O. Box 500, Beaverton, Oregon 97005.
New ways in digital automation with:

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- Synchro-to-Digital Converters, 6 types
- Analog-to-Digital Converters, 10 types
- Solid-State Digital Modules, 136 types

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Selecting from hundreds of off-the-shelf digital transducers and circuits, we assemble a remarkably low-cost, customized digital system for control/display/logging. DECITRAK can perform as simple a job as providing remote digital display of antenna pedestal position. Or, as impressive a task as the precise control of massive prime movers in response to punched-card commands.

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Hermetically-sealed in metal cases. Four case sizes, ranging from 3/4" to 1/4" length. Value-packed performance characteristics — low impedances at high frequencies, low dissipation factor, minimal capacitance drift with temperature, practically no change in capacitance with life. Low leakage current limits. Investigate new higher capacitance ratings.

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Vitreous-enamel BLUE JACKET® POWER WIREWOUND RESISTORS

All-welded end-cap construction eliminates moisture along the leads, also anchors leads securely to resistor body. Expansion coefficients of vitreous enamel coating, ceramic core, and end caps are closely matched. Standard wattage ratings include 1, 2, 2.5, 3, 5, 7, 10, and 11 watts. Also available with radial tab terminals in ratings from 8 to 230 watts.


18
'Big Horn' antenna at Andover, Me., may one day receive and transmit messages from a domestic satellite system. It's now used as an Intelsat ground station. p. 36.

365-foot solar telescope will attempt to unlock the sun's secrets. p. 30.

Transmission of data from Mars at 16,000 bits per second became feasible when a 210-foot antenna went into operation. The result: best photos yet taken of the planet. p. 25.

Also in this section:

Animal tests to help evaluate TV radiation. p. 32.
Burroughs, the originator of NIXIE® tubes, now revolutionizes display technology with the first commercially practical dot matrix display system. It took SELF-SCAN panel display, the remarkable Burroughs invention that takes the electronics out of the present electronic display ... reducing costly drive circuitry up to 90%.

With a minimum number of leads and drivers, Burroughs' system automatically scans data input into in-plane readout characters formed by glowing dots ... making possible a totally new combination of readability, minimal packaging and cost advantages.

The new flat-panel display is basically a matrix of small gas discharge cells hermetically sealed between heavy glass plates in a sandwich configuration. The matrix itself, formed of insulating material, has small grooves on its top surface to allow positioning of information anode wires which intersect each hole. Cathode conductors behind the center sheet intersect at each cavity with a second set of anode wires.

By utilizing the phenomena of preferential glow transfer and glow shifting, the initial glow caused by cathode ionization in the dot matrix may be moved through selected holes to a visible position on the top surface. A sequential flow of light is thus achieved without separate drivers for individual columns and rows.

Burroughs' alphanumeric SELF-SCAN panel displays are available with or without memory for any application requiring 16 or 18 digit readout.

Write today for descriptive brochure, Burroughs Corporation, Electronic Components Division, P. O. Box 1226, Plainfield, New Jersey 07061. Tel.: (201) 757-5000.
Steady growth forecast for electronics in '70s

Despite expected cuts in spending by the Defense Dept., the outlook for the electronics industry in the early 1970s is good, the Electronic Industries Association says.

In a report to be presented at its meeting in Los Angeles Oct. 6-8, the EIA says electronics sales should rise from $25.1-billion this year to $26-billion by 1971-72. Total defense spending will be the EIA says electronics sales to rise from $25.1-billion this year to $26-billion by 1971-72. Defense spending will again rise, after 1972, the EIA forecasts, during the later 1970s.

No single market continues to dominate electronics sales, according to the association. But it was emphasized that just in what direction and how much should be spent for new market development is entirely up to the individual firms, since the EIA report makes no recommendations.

The projected decrease in defense spending during the next three years is based on a slowdown in the Vietnam war effort, primarily in such nonelectronic areas as ammunition outlays, personnel and operations and maintenance. These decreased costs will be partly offset by inflation and increased spending for strategic systems such as a new aircraft to replace the B-52 intercontinental bombers.

The marketing report gives a comprehensive review, in 13 chapters, of the many areas of interest to various segments of the electronics industry. These include 1-year projected studies for such areas as aerospace, law enforcement, marine sciences, poverty, housing, transportation, and the balance of payments situation.

To those who look for rapid expenditures in such areas as pollution, crime control, urban reconstruction, transportation and education, the report's predictions will be disappointing. Not before five years, or possibly 10, will the full impact of spending in these problem areas be felt, the EIA says.

An analysis of the market in oceanography indicates that expenditures will stay at the half-billion dollar annual level for the near future, with an upturn to about $1.8-billion by 1980.

The study invisions: "This turning to the sea and the prospect of a national program directed a fostering civil marine technology have major implications for the aerospace and defense industry."
News Scope CONTINUED

tions, a tower and an instrument landing system.

Latrobe didn’t have these and the Federal Aviation Administration couldn’t help because the airport didn’t have enough landing operations to warrant the government’s supplying the equipment.

So, Latrobe bought the equipment itself. The Pennsylvania Aeronautics Commission kicked in half the money, and corporations and private donors provided the rest. They bought an ILS from Cutler-Hammer’s AIL Div. in Deer Park, N. Y., for $79,000. A trailer-like tower, hoisted to the roof of a building was obtained from Air Traffic Control Systems, Inc., in Cleveland for about $25,000.

Since then monthly operations have jumped from 1500 in October 1968—just before the new equipment was installed—to 8240 in August 1969. The sale of gasoline has increased 67 per cent, and landing fees are rolling in. The airport is so busy that FAA will probably soon take over its operation.

Latrobe airport authorities are happy, and electronics manufacturers are, too. AIL has sold landing systems for non-FAA money to three other towns and is looking optimistically to the 300 additional community airports throughout the country that also might try it alone.

Stock exchange makes automation study

The American Stock Exchange has released a broad-based blueprint for computerizing security-exchange procedures and operations from coast to coast. The study, prepared for AMEX by North American Rockwell Corp., took six months to prepare. It examines the exchange’s problems, procedures and operations and recommends what should be automated and what left alone.

“We have task forces looking at hardware,” an exchange spokesman said, “but any selection is a long way off.”

Another study, looking farther into the future, is under way by the Rand Corp. under contract jointly with both the American and New York Stock Exchanges. This study will be completed by February, 1970.

New vendor offers line of precision ICs

Unusually precise analog ICs will be manufactured by Precision Monolithics Inc., a new company of Santa Clara, Calif., according to its president, Marvin Rudin.

“One of our first products will be a monolithic 741 op amp with offset currents under 3 nA and offset voltages under 1 mV,” says Rudin. “This will be an improvement by a factor of at least 10 in current and 5 in voltage over the best 741 now available from other sources.”

The op amp and other circuits will compete directly with the high-performance linear modules produced by a number of vendors, and will include DDAs, opamps, and precision linear amplifiers. The company expects to fill initial orders by the first quarter of 1970.

Precision Monolithics Inc. is 40% owned by Bourns Inc., Riverside, Calif.

RCA plans expansion in time-sharing field

For several years, RCA has been dangling its corporate toes in the soothing, tepid waters of commercial time-sharing without taking the plunge. Now, it has indicated, it may be ready to dive into the market.

The occasion was the dedication earlier this month of RCA’s plant in Marlboro, Mass., where it plans to turn out peripheral computer equipment. Robert W. Sarnoff, the company’s president, announced at the ceremony the introduction of the RCA Spectra 70/61 computer—an advanced and sophisticated machine designed for use in the time-sharing field. RCA has every intention of using the new computer to go on line with a wide range of time-sharing services, Sarnoff said.

Both Sarnoff and James R. Bradburn, executive vice president, pointed out that RCA currently offers shared computer time to companies in Wall Street and also has its own internal scientific time-sharing service as a link between RCA plants across the country.

Important features of the new 70/61 computer include the ability to provide local and remote access for over 350 terminal users. It is fully compatible with the earlier and smaller 70/46 and has source-level language compatibility with the IBM System 360. Monthly rental for the 70/61 will be about $40,000, including maintenance. This price is 30 per cent higher than for the 70/46, but the larger machine is three times faster, RCA says, and also offers a threefold increase in terminal capacity and a fivefold rise in input/output capacity.

Sarnoff said: “During the 1970s we expect RCA’s commercial data-processing equipment sales to rise beyond $1-billion a year.” He added:

“By 1975, 80 cents of the average dollar spent on systems will be for the kinds of peripheral equipment to be manufactured at Marlboro.”

Weather center adds new computer system

Meteorological data analysis and prediction got a boost with the acquisition, last month, of a second Control Data Corp. model 6600 large-scale digital computer system by the Commerce Dept.’s Environmental Science Services Administration (ESSA). The 6600 was installed at the World Weather Center at Suitland, Md., a suburb of Washington, D. C.

The facility is shared by the National Environmental Satellite Center and the National Meteorological Center. The former agency operates the Tiros Operational Satellite weather system, which continually receives data from the Environmental Survey Satellites in orbit around the earth. The Meteorological Center collects and analyzes data from all global sources.
Fairchild Acquires Union Carbide Modular Products Line.

All standard and special modular op amps formerly offered by Union Carbide are now manufactured and marketed by Fairchild Controls, Mountain View, California.

Fairchild Controls' purchase of these Union Carbide products is part of a continuing planned program of expansion through both internal new product development and acquisition of quality products which complement the Fairchild line.

Fairchild will also continue to produce all special modular op amps that until now have been provided by Union Carbide.

For availability or applications information regarding these products contact Fairchild Controls, Modular Products, 423 National Avenue, Mountain View, California 94040. (415) 962-3833.

Standard Union Carbide operational amplifiers now made and sold by Fairchild Controls include:

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<thead>
<tr>
<th>Class</th>
<th>Type</th>
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<td>General Purpose</td>
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<td>H6050</td>
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<tr>
<td>Comparator</td>
<td>General Purpose</td>
<td>H8010</td>
</tr>
</tbody>
</table>
Looking for an All-Around Voltmeter?

Hewlett-Packard gives you a broad choice of multi-function meters that do not have to be pampered. Choose the versatility that fits your needs for ac volts, dc volts, current and resistance measurements. The exclusive individually calibrated taut-band meter used in these voltmeters gives you reliability, repeatability and high accuracy.

Make 90% of your day-to-day ac/dc measurements with laboratory precision using the hp model 410C Voltmeter. Measure dc from 15 mV to 1500 V full scale, current from 1.5 µA full scale, resistances from 10Ω to 10 MΩ, and ac volts to 700 MHz. The hp-developed photoconductor chopper amplifier gives the 410C high sensitivity, low drift, and low noise. Price of hp 410C is $475. Vacuum tube version, hp 410B is $300.

Low cost fully-portable multi-function meter—that's the all-solid-state, battery-operated hp model 427A Voltmeter. It costs only $225. Option 01 gives both battery and line operation for an additional $25. Measure dc voltages from 100 mV to 1 kV full scale; ac voltages from 10 mV to 300 V full scale at frequencies to 1 MHz (to 500 MHz with the 11096A High Frequency Probe, price $45); resistance from 10 Ω to 10 MΩ. Ac and dc accuracy is ±2%. FET's in the input circuit give you 10 MΩ input impedance—minimal circuit loading.

Highly sensitive dc and resistance measurements are made with hp 412A DC Vacuum Tube Voltmeter. With its 1 mV full scale dc voltage sensitivity and 1 Ω midscale ohms sensitivity, and its simplicity of operation, the 412A is ideal for production line testing. Measure dc with 1% full scale accuracy. Price of 412A is $450.

Extreme accuracy and hands-free operation distinguish the “Touch and Read” 414A DC Autovoltmeter. Automatic ranging and polarity indication occurs in less than 300 ms. Measuring accuracy for dc voltage is ±(0.5% of reading +0.5% of full scale)—the best available in any analog voltmeter. Resistance accuracy is ±(1% of reading +0.5% of full scale) on an easy-to-read linear scale. Price is $690.

For full details on these and other Hewlett-Packard Voltmeters, see your hp catalog or contact your nearest hp field engineer. Or, write to Hewlett-Packard, Palo Alto, California 94304. Europe: 1217 Meyrin-Geneva, Switzerland.
From 8 1/3 bps to 16,200 bps--Here's how

New telemetry for Mariner began as experiment and ended with the best pictures yet taken of Mars

David N. Kaye
West Coast Editor

It began as an experimental high-speed telemetry system that Mariners 6 and 7 were to test as the unmanned spacecraft flew past Mars. But it worked so well in the developmental stage that NASA decided to make it the primary transmission system on the space missions last July.

As a result, data was transmitted back to earth at a speed of 16,200 bits per second instead of the 270 bps originally planned. And NASA obtained TV photos of Mars that were four times better than those obtained in the first Mariner fly-by in 1965 (see ED 17, Aug. 16, 1969, "New Mars Photos Show a Fourfold Gain in Detail," p. 21).

The significance of the new high-speed data-transmission system becomes clear when one considers that the 270-bit-per-second system originally planned for Mariners 6 and 7 was already quite a jump from the 8 1/3-bps system used in 1965. Instead of being the primary system, the 270-bit system was used as a secondary transmission unit.

Recorder is bypassed

With a primary data-transmission speed of 16,200 bps, NASA was able to bypass one of the two on-board tape recorders when the Mars scientific and picture data were sent back to earth. For certain data, both recorders were bypassed. Not only could a greater amount of data be sent back in a shorter time, but increased system reliability was achieved through the elimination of one or both of the tape recorders.

Original plans called for all the data to be stored on board the spacecraft at high speed on an analog tape recorder, then played back at a lower speed through an analog-to-digital converter, and finally re-recorded on a digital tape recorder. A playback of the digitized tape at a still lower speed was then planned so the data could be transmitted to earth over the 270-bps telemetry channel. Engineers at Jet Propulsion Laboratory in Pasadena, Calif., recognized, however, that the use of two tape recorders in tandem created a potentially unreliable situation. A question was raised as to the feasibility of using a special channel that would transmit the data at the playback rate of the analog tape recorder; this would eliminate one of the tape recorders. At 16,200 bps, it was noted, the analog tape recorder could play back all of the recorded data in slightly less than three hours. Subsequently it turned out that for certain data, even the analog tape recorder could be eliminated.

According to Mahlon Easterling, project engineer on the high-speed telemetry system in its early stages at JPL, "We could play back scientific data and decimated picture data as it was acquired. Decimated picture data includes every seventh picture element." To get back all the data in real time would have required seven times the speed that was available, or 113,400 bps. And to transmit fast enough for real-time motion TV, the power would have had to be increased a great deal more than seven times. NASA settled for the 16,200-bps system.

"Since the science data and the decimated picture data could by-

A key ground factor in the success of the Mariner 6 and 7 missions was this 210-foot antenna at Goldstone, Calif. It wasn't available for the 1965 flight.
Telemetry, continued

pass the recorders,” Easterling observes, “it could come back with a higher quality. Each recorder does degrade the data somewhat.”

Several factors contributed to successful increasing of the transmission speed to 16,200 bps. The most important was a switch from an 85-foot antenna on the ground to a 210-foot dish put into service at Goldstone, Calif., in 1966.

Another important move included increasing the spacecraft’s transmitter power from 8.9 to 18.2 watts, and transmitting the data in a special code.

Easterling notes: “Each bit requires a specific amount of energy. Since power is energy per unit time, doubling the power, doubles the number of bits per unit time which can be transmitted.” But each factor must be related to the total energy received on the ground.

To send 16,200 bps instead of 8 1/3, an increase in power of 32.9 dB had to be provided. Most of this came from the factors that make up the standard communication equation:

\[ S = P_T M G_T L_s G_B L_B \]

in which \( P_T \) is the power produced by the spacecraft transmitter; \( M \) is the modulation loss (that is, the factor that relates the power in the sidebands to the total power); \( G_T \) is the transmitting antenna gain over an isotropic antenna, including both circuit and pointing losses as well as antenna gain; \( L_s \) is the free space transmission loss; \( G_B \) is the receiving antenna gain; \( L_B \) is the receiver loss or efficiency (the difference between how the receiver actually operates and how it would operate if perfect), and \( S \) is the received sideband power.

The received sideband power multiplied by the duration of a bit, \( T_B \), gives the received signal energy per bit. The receiver system effective noise temperature multiplied by Boltzmann’s constant gives the receiver noise spectral density. Together, these specify the figure of merit, \( ST_0 / N_0 \), for a digital communication system. Dividing the achieved figure of merit by that required for a specified performance, we get the system’s margin of performance.

The 32.9 dB of additional signal was obtained in the following way.

- 3.10 dB was due to an increase in the transmitter power, \( P_T \), from 8.9 to 18.2 watts.
- 3.96 dB was due to a change in the modulation index, \( M \), which increased the sideband power from –5.30 to –1.80 dB relative to the total power.
- 0.10 dB was due to an increase in the size of the spacecraft antenna. Although \( G_T \) should be substantially larger, a compromise was made in which the antennas on both spacecraft were pointed in the same direction, even though they arrived near Mars on different days.
- 6.84 dB was picked up in \( L_B \)

The payoff of a faster telemetry system is readily apparent when these two photos of Mars are compared. Left: A shot by Mariner 4 in 1965, transmitted at 8 1/3 bits per second. Right: A sharper, more detailed photo of the distant planet sent last July by Mariner 7 at 16,200 bits per second.
Signetics Corporation / 811 E. Arques Ave., Sunnyvale, Calif. 94086 / A subsidiary of Corning Glass Works

INFORMATION RETRIEVAL NUMBER 16

Electronic Design 20, September 27, 1969 27
due to the decreased distance between Mars and the earth, as compared with the 1965 Mariner voyage. This time around, Mars was less than half the distance from the earth than it was four years ago.

- 8.50 dB was due to the increased antenna gain on the earth. $G_e$ for the 210-foot antenna at Goldstone is +61.00 dB, as opposed to +52.5 dB for the 85-foot antenna used before.

- 2.96 dB was due to improved receiver efficiency, $L_R$, on the ground. According to Easterling: "The equipment is about 90 per cent efficient—that is, it operates within 10 per cent of what it would if it were all perfect. This is perfect in the ideal, or block-diagram sense."

These factors accounted for 25.46 dB of the required 32.9 dB. Most of the remaining improvement came from two other factors:

- 4.10 dB was due to the fact that Goldstone has a maser that provides a system noise temperature of 25° K instead of the 65° K that the earlier system had. After conversion to noise spectral density, $N_0$, the new figure became -184.60 dBm/Hz.

- 2.20 dB was due to the block coding used. If bit-by-bit detection were used, as in Mariner 4 in 1965, the required $ST_{N_0}/N_0$ would have been 5.20 dB after all losses were considered. But with the data encoded six bits at a time into a biorthogonal code, the required $ST_{N_0}/N_0$ was reduced to only 3 dB, for a word-error rate of $10^{-6}$. That error rate was equivalent to the error rate on Mariner 4.

With 31.76 of the 32.9 dB now accounted for, the last 1.14 dB came in the form of a reduction in the performance margin—from 3.30 dB on Mariner 4 to 2.16 dB on Mariners 6 and 7.

Reed-Muller code used

The heart of the high-rate telemetry system was the block coder, built for JPL by Texas Instruments in Dallas. It could take a six-bit data word and transform it into a 32-symbol, block-coded word. The result was data-coded with the use of a form of Reed-Muller biorthogonal code.

The final output was also comma-free, so that word sync could be derived on the ground. Comma freedom in a code means that any 32-symbol string composed of the last 32-N symbols of one code word in juxtaposition with the first N symbols of the next code word (that is, an overlap of N symbols) is forbidden as a member of the code dictionary. Therefore from any string of symbols received on the ground, a definitive determination could be made of where one code word ended and the next began.

The block coder took either the real-time data or the stored data, converted it to code and fed it into a 180-degree phase modulator. The phase modulator placed the data on a 269.2-kHz square-wave subcarrier and sent the result into a summing circuit. In the summer, a 24-kHz engineering telemetry signal was added, and the sum was fed into another phase modulator. Here the information was placed on a 2295-MHz, S-band carrier, and the final signal was sent to the transmitter. The transmitter increased the power to 18.2 watts and fed the signal through a 40-inch, high-gain antenna and back to the earth.

"The major problem in the development of Mariner '69," according to Easterling, "was in the precise testing of the system. For example, how could we establish within a couple of tenths of a dB the precise S-band signal levels at which we had to work."

Test console a big help

James C. Springett, the engineer at JPL in charge of verifying the high-rate telemetry system performance, credits an rf test console built by Westinghouse with solving the most serious problems. This test console is capable of creating a signal source with a known signal-to-noise ratio. Thus the system under test sees the signal-to-noise ratio that it will see in actual performance. The console has a $\pm 0.1$ dB/day stability and a $\pm 0.05$ dB accuracy, for an over-all accuracy of $\pm 0.15$ dB. According to Springett, this is an "order of magnitude better accuracy" than that obtained with normal techniques.

High-data-rate telemetry is also planned for a Mariner probe scheduled in 1971. Two vehicles that year will orbit Mars, taking pictures and sending back scientific data. The basic data rate is to be the same as that for the two recent Mariners.

"The main evolution on Mariner '71," says Easterling, "will be to have multiple data rates available for use."
2.5A Triac Versatility from RCA

Sensitive-gate Triacs in TO-5 package for low-power phase control and load switching

Here's versatility plus in a Triac family! RCA types 2N5754, 2N5755 and 2N5756 are designed for low-current circuits of either resistive or inductive loads, and for use in driving high current Triacs. All three devices feature shorted-emitter design and RMS on-state current rating of up to 2.5 A. All have gate characteristics as follows: $I_{gs} = 25$ mA max. in $I^+, III^-$ modes and 40 mA max. in $I^-, II^+$ modes. Repetitive peak off-state voltage ratings of 100, 200 and 400 volts are available to match your needs. And all three Triacs are designed to switch from off-state to on-state for either polarity of applied voltage with positive or negative gate triggering voltages.

For PC board installation, you can also obtain the 2N5754, 2N5755 and 2N5756 with factory-attached heat-radiators. Ask for RCA's 40684, 40685 and 40686, respectively.

You can also turn to RCA if you need greater gate sensitivity. The 40530-series Triacs offer you $I_{gs}$ of 3 mA and 10 mA—for driving by integrated circuits, as shown in the following tabulation:

<table>
<thead>
<tr>
<th>$I_{gs}$</th>
<th>Low Voltage (100 V)*</th>
<th>120-V Line (200 V)*</th>
<th>240-V Line (400 V)*</th>
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<tbody>
<tr>
<td>3 mA (all modes)</td>
<td>40525/40531</td>
<td>40526/40532</td>
<td>40527/40533</td>
</tr>
<tr>
<td>10 mA (all modes)</td>
<td>40528/40534</td>
<td>40529/40535</td>
<td>40530/40536</td>
</tr>
<tr>
<td>25 mA ($I^+, III^-$ modes)</td>
<td>2N5754/40684</td>
<td>2N5755/40685</td>
<td>2N5756/40686</td>
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<tr>
<td>40 mA ($I^-, II^+$ modes)</td>
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</tbody>
</table>

* $V_{CRM}$ (Repetitive peak off-state voltage)

Select the type best suited for your circuit designs. Then see your local RCA Representative or your RCA Distributor for economical pricing details. For technical data, write RCA Electronic Components, Commercial Engineering, Section R034, Harrison, N. J. 07029.
Computerized ‘eye’ to shed light on the sun

16 pre-programmed investigations will be available when Air Force puts new solar telescope to work

John F. Mason
Military-Aerospace Editor

When the world’s most sophisticated solar telescope takes its first look at the sun next month, Air Force scientists atop the 9200-foot mountain near Sunspot, N. M., will have little to do but sit back and watch the secrets that the gigantic instrument reveals (see “Giant Telescope to Probe Sun’s Secret,” ED 19, Sept. 13, 1969, p. 22).

With a resolving power of less than 0.2 seconds of arc, the telescope’s acuity matches that of the most powerful telescopes—stellar or solar—in the world.

The new instrument will surpass the performance of the previous champion—the solar telescope at Kitts Peak, Ariz.—in clarity and resolution. Although the Kitts Peak instrument is longer—500 feet, as opposed to the new telescope’s 365 feet—the new instrument is evacuated, thus eliminating air currents that distort the image.

The operation will begin when a sensor tells the telescope’s SDS Sigma 2 computer that the sun is out and ready to be observed. On this cue, the computer will take over, putting the 365-foot instrument through one of 16 pre-programmed investigations of some aspect of the sun that the scientists select.

A 321-foot-long vacuum chamber will enable this solar telescope to provide high-resolution, jitter-free pictures of hitherto unseen solar activity.

The SDS Sigma 2 is built by Scientific Data Systems, a subsidiary of Xerox Corp. at El Segundo, Calif. It is a 24-kilobit, 900-nanosecond machine.

The telescope will be used to examine the solar corona—the most difficult region of the sun to observe—as well as flares and prominences, sunspots and the high energy proton showers associated with sunspot activity. Knowledge of these phenomena will help scientists from the Air Force Cambridge Research Laboratories predict severe magnetic and other solar disturbances and develop methods for minimizing their effects. These phenomena have long been a menace to communications on earth, to instruments in satellites, and, now, to man himself as he contemplates traveling deeper into space.

The scientists can select the particular phenomenon they want to watch by flipping one of 16 switches on a hand-held control box. Eight of these boxes will be in the control room.

Computer calls the shots

The computer’s first step, after being informed that the sun is out, is to command the telescope’s turret to find the sun. When this has been accomplished—roughly within a minute or two of arc—the computer tells the turret’s servo control system to switch to its optical error detector mode, to align the turret precisely with the sun and then to track it as it moves.

At the same time the computer starts the coordinate converter, which converts the azimuth and elevation angles of the turret to astro-geometrical data. This information, along with a solar clock, enables the telescope to track the sun while it is obscured by clouds.

Information from the telescope is picked up by a photomultiplier and put on magnetic tape. The first readout appears on microfilm in the
form of plots and graphs.

Of the telescope's 365-foot length, 227 feet is set in a deep shaft beneath the ground. The above-ground portion is enclosed in a conical tower 138 feet high. The central core of the telescope consists of a cylinder 321 feet long with a maximum diameter of 10 feet. The system's entire optics—from the objective port at the top of the tower to a 64-inch reflecting mirror at the bottom of the shaft—is in an evacuated chamber.

Eliminating distortion

The chamber weighs 250 tons and is evacuated to 250 torr, a pressure that corresponds to that at an altitude of about 180,000 feet. To provide a friction-free bearing for the entire core assembly, the chamber is suspended like a pendulum, floating on an 11-ton pool of mercury. The assembly rotates slowly to follow the sun.

Besides evacuation of the chamber to eliminate distorting internal air currents, additional protection against image-distorting heat currents is provided by embedding water-cooled pipes in the concrete walls of the tower. By control of the temperature of the water flowing in these pipes, the temperature of the tower walls can be maintained at equilibrium with the air temperature inside the tower.

Light from the sun enters the telescope through a 76-cm aperture via a quartz window 10 inches thick and 30 inches in diameter. Two flat mirrors, each 40 inches in diameter, reflect the light—first along a horizontal path, then downward through the 321-foot length of the cylinder to a focusing mirror 183 feet beneath the ground level. This 64-inch focusing mirror has a focal length of 180 feet. It can be titled to direct light upward again and to form an image of the sun on one of the several selected instruments at ground level.

All of the instruments are clustered about the central shaft and rotate with the shaft. The largest instruments are three evacuated spectrographs, each 68 feet long with diameters of five feet. These are mounted vertically to the central core.
Animal tests to help evaluate TV radiation

How harmful—if at all—are the effects of radiation from color TV sets? Pigs may provide some clues. The Dept. of Health, Education and Welfare is conducting a series of experimental studies in which the eyes of swine are being irradiated to compare the effects from color TV radiation with those from high-energy X-rays.

According to a spokesman for the federal department, pigs are being used because, "except for subhuman primates, their optical system is closest to that of humans." The experiments are being conducted at the department's Bureau of Radiological Health in Rockville, Md.

The focus of the research, according to Harry D. Youmans, physics adviser to the bureau, "is to compare known biological effects from 250-kV radiation with 30-kV, color-television-like radiation."

The effects suffered by pigs subjected to 100, 600 and 1000 Rads of 30-kV X-rays cannot be compared directly to the effects on humans, he says, adding: "But these levels are not as far-fetched as they may seem. Information we have received indicates that some children watch TV for as long as five to 15 hours a day. And the total radiation from color TV sets—if the set is defective—can be quite high."

Several surveys of radiation from color TV sets have been conducted by the Public Health Service. In Suffolk County, N.Y., officials from the PHS found that 20 per cent of the sets they examined emitted radiation above the 0.5-mR/hr level. A review of the survey taken in Pinellas County, Fla., shows that about 65 per cent of the sets emitted radiation above 0.40-mR/hr.

Speaking for the TV manufacturers, J. Edward Day, special counsel for the Electronic Industries Association denies that a hazardous situation exists and that there is any justifiable cause for public alarm.

Day told ELECTRONIC DESIGN that the Suffolk and Pinellas County surveys "are unreliable and that the measuring instruments were inaccurate."

Day considers a survey conducted by the PHS in Washington, D.C. to be a better reflection of the true situation. In this survey, about 6 per cent of the sets were found to exceed the 0.5 mR/hr standard. However, two out of the 1128 receivers studied emitted radiation above 12.5 mR/hr—the top of the measuring scale.

The reason for ionizing radiation emissions is the operating voltages for color sets, which is about 10 times greater than the black and white receivers. Sources of this radiation can be eliminated by going from tube components to solid-state devices.

The Dept. of Health, Education and Welfare has set down the following guidelines for permissible radiation from manufacturer's sets:

- Effective Jan. 1, 1970—when power-line voltages are at the highest level, radiation must not exceed 0.5 milliroentgens per hour at a distance of 5 cm, averaged over an area of 10 cm.
- Also, by Jan. 1, 1970—sets must pass this test with 130-volt input and all user TV controls set to positions producing the maximum radiation.
- By June 1, 1970—sets must pass this test with all service controls set to produce maximum radiation.
- As of Jan. 1, 1971—measurement must meet the 0.5-mR/hr standard, even with defective, shorted or inoperative components.

The 0.5-mR/hr standard adapted by HEW is a figure that has been set by the National Council on Radiation Protection and Measurements. A spokesman for that organization says there has been no definitive study of the physical effects of low-level radiation and that the 0.5-mR/hr standard is conservative—"less radiation than that experienced by persons living in, say, Denver."

George M. Wilkening, head of industrial hygiene and safety at Bell Telephone Laboratories and a member of the Technical Electronic Product Radiation Safety Standards Committee set up by the government, also points out that "while relatively low-level radiation can and should be reduced, there has been no scientific information on the effects such radiation may have on the body."

This is why the research at HEW's Bureau of Radiological Health is important. Their researchers may find some answers to the biological effects of low-level radiation—in a pig's eye.
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This new Cunningham Hystareed Magnetic Latching Relay operates without permanent biasing magnets or holding currents. Latches and stays latched. Won't change its programmed state because of vibration, shock, power interruptions or transients. Needs no adjustment or calibration. Available up to 6 form A's, 4 form C's in a single assembly. Operates as standard relay for scanning.

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CORNING ELECTRONICS

INFORMATION RETRIEVAL NUMBER 20
What's delaying U.S. satellite communications?

The technology is ready and able. Only a decision from Washington is needed.
Looming up out of a field of cane at Cayey, Puerto Rico is this 85-foot earth station antenna. The station links the island with the U.S. mainland, Latin America, Europe and other Atlantic Basin points equipped with earth stations.

public-owned utility to develop and operate the U.S. portion of an international satellite telecommunications system. Subsequently Comsat was selected by the Satellite Telecommunications Satellite Consortium (Intelsat) now numbering 68 nations, to manage the global network.

U.S. common carriers and other interested industry groups accept Comsat's role in the international system but deny that it has the automatic right or authority to own and operate all domestic systems. They recognize the vast potential for TV, telecommunications, educational broadcasting and aeronautical communications and navigation support. The clash over who will operate a domestic satellite system is further tangled with future ownership of ground terminals.

Comsat believes it has a Congressional mandate through the act that established it as a "carriers' carrier." James McCormack, Comsat's board chairman, sums up the corporation position this way:

"In Comsat's view, it would be entirely without logic or economic sense to contend that the nation's only presently authorized operator of privately owned satellite facilities was ever intended to be excluded from the nation's first really large market for satellite communications. We are more than willing to provide all needed satellite services to all potential users, foreign and domestic. We have no other reason for being."

AT&T agrees that Comsat should retain control of satellite operations but says it should not own all supporting ground stations. Rather, terminals should be owned and operated, where appropriate, by the common carriers, AT&T says.

Not so, says Comsat's McCormack. Ownership of earth stations is "vital" to Comsat, he declares, since the young corporation "has no interest in any other communications medium and no other means of diversification."

Comsat is not opposed by just the U.S. carriers. Many foreign governments, members of Intelsat, argue that Comsat participation in a U.S. domestic system could constitute a conflict of interest. The reason: Comsat is the U.S. representative in Intelsat and controls the majority vote in all decision-
making processes; also, Comsat is the operational manager for the Intelsat global network.

Meanwhile, to get the system moving, Comsat has offered to begin establishment of an experimental domestic system using its own money. Later, if the FCC should so determine, it would turn the system over to another designated operator (with proper reimbursement).

Congress has held hearings with interested parties in an attempt to better understand the problem and determine whether or not additional legislation is necessary. To date it has found no such need.

President Lyndon B. Johnson established a task force in 1967 to develop recommendations for future U. S. telecommunications policy including satellite techniques. The group's report, completed last year, strongly urged initiation of a pilot or experimental domestic satellite communications system. The report, only recently released to the public by the next Administration, was put out without any endorsement of the findings. The FCC has deferred any decision on domestic satellite policy with the excuse that it would await the task-force report recommendation on future policy and, hopefully, some expression of direction from the President.

**Further delays sought**

The White House stepped in last month by requesting further delay (at least 60 days) of a decision until Presidential Special Advisor on Communications Dr. Clay T. Whitehead and his staff can review the domestic satellite system problems and industry proposals. It has also requested FCC cooperation and participation in the Whitehead study.

Ultimately, the FCC commissioners must weigh all recommendations and determine some course to follow. Their problem is further muddled, in the opinion of many experts, by the fact that the FCC may not have the legal power to decide the issue. Nevertheless, it is expected that a policy determination will be made shortly, though the real solution may come later in the courts.

On Dec. 16, 1966, Comsat filed a proposal for a domestic communications satellite system and service with the FCC. It contained a comprehensive outline for an initial or experimental system but involved a two-phase program that included three generations of hardware. In short, the Comsat proposal was based on a long-term building-block approach in which experience can contribute to design improvement as the system grows.

The initial phase, called the Pilot Program, proposed a two-year effort during which up to four high-capacity multipurpose synchronous satellites, each carrying 12 transponders, would be deployed. These would operate in the 5925- to 6425-MHz band for receiving (up-link) and in the 3700- to 4200-MHz band for transmitting (down-link).

A total of 34 ground stations would be needed initially, including two extra-large stations in the New York City and Los Angeles areas. The large terminals would each employ two 85-foot antennas. Two other receive-transmit stations located in the southeastern and northwestern regions of the U. S. would use single 42-foot dishes. The remaining 30 smaller terminals would be for receiving only and would use dishes from 25 to 35 feet in diameter, depending on their locations.

Each satellite would be capable of handling 12 color TV channels, or 21,600 trunk-message channels (through the two 85-foot dishes only), or 9600 multipoint message channels (to 42- and 85-foot dishes only). All stations would support TV receiving.

For the Pilot Program, Comsat estimates a space-segment investment of about $36 million and an additional $22 million for the ground segment. It also estimates a two-year operational cost of about $47 million.

The design of the present Intelsat IV satellite (now in development by Hughes Aircraft) will fit the requirements for the Pilot Program vehicle, in addition to serving the global Intelsat network as fourth-generation hardware.

Comsat emphasizes that, even with the early domestic system, it would be able to provide reliable and efficient 24-hour multi-channel transmissions to all time zones of TV network programs for distribution to affiliate broadcast stations. Its proposal also asserts that the system will provide other "cultural and educational TV distribution, trunk and point-to-point telephone, telegraph, facsimile, data and other communications."

Ultimately, in its "1970 model" system, four satellites could be employed, each with a five-year lifetime. These would provide a total capacity of 48 color TV channels or up to 84,000 trunk voice-grade channels, or any combination of these. Earth stations would continue to be expanded so that, by the time all four satellites were in service, up to 180 ground stations would be operational. The firm estimates that by about early 1971 90 per cent of the U.S. TV-viewing public could be receiving satellite-relayed programs.

The Comsat technical plan next calls for four "1973 model" satellites, each having double the capacity of the earlier model and each having a six-year design life. This model, too, would operate in the 4- and 6-GHz bands but would also be capable of receiving at frequencies above 10 GHz. The more advanced satellite array, when fully deployed, would be capable of handling 96 color TV channels, or 168,000 telephone channels.

**AT&T seeks growth system**

Through its subsidiary, Bell Telephone Laboratories, AT&T also proposed in detail, late in 1966, a two-phase domestic system using multipurpose satellites and a very large operational capability.

AT&T noted that it seeks only to own and operate the ground station along with other U. S. common carriers. The satellite portion of the system, says AT&T could be owned and operated by Comsat.

For its initial system, which would cost more than $290 million, AT&T suggests a concept similar to Comsat's proposed Pilot system. It would use two stationary satellites operating in the 4- and 6-GHz bands, with a capacity of either 12 TV channels or 9600 duplex voice circuits. Two major Earth stations would be needed, using 85-foot dishes—one in the New York area and one in the vicinity of Los Angeles. In addition, 73 TV receiving stations.
would be required for national distribution of programs.

The second phase, however, involves new ground stations and highly advanced satellites to come into service in the early 1970's. The satellites would be large, on the order of 3000 pounds and would operate in the bands of about 4, 18 and 30 GHz.

The second-phase system would probably cost more than $200 million, the firm estimates. This would include four satellites and their launch vehicles, plus extensive modifications to the existing ground facilities. But the firm is anticipating an expected spectrum congestion in the 4- and 6-GHz bands. It further believes that satellites operating at these lower frequencies will not have sufficient traffic-handling capability to meet traffic needs anticipated by the mid-1970's.

The larger satellites each would have forty-five PCM (pulse code modulation) and 12 fm transponders. Each PCM transponder would be capable of handling 1344 simplex telephone circuits or one TV circuit; each fm transponder would handle one TV channel. The fm transponders would operate at the 4-GHz level while the PCM transponders would receive at 30 GHz and transmit down at 18 GHz.

AT&T estimates that 26 ground stations would have to be available for the advanced satellites at a total cost of nearly $70 million.

Total capacity of the system by the mid-1970's would be 83,000 voice circuits and 27 TV circuits. This should be sufficient, the firm says, until about 1980.

Ford wants educational TV relay

In mid-1966, the Ford Foundation submitted a proposal to the FCC for what it termed a "broadcasting non-profit satellite system." Essentially it would be used as an educational TV relay system to reach all areas of the continental U.S., as well as Alaska, Hawaii, Puerto Rico and the Virgin Islands. The multichannel system would have a high-quality TV color channel for each of 24 satellite microwave repeaters.

In the Ford concept, two satellites would be placed in synchronous orbit, and a third might later be considered for increased traffic capacity. The satellite design concept, prepared by Hughes Aircraft and Philco-Ford, calls for a craft weighing in at about 770 pounds and using an electronically despun antenna system.

The ground system would require up to 233 terminals of varying complexity for metropolitan areas, plus 10 mobile stations.

The last of the major proposals submitted to the FCC for consideration in the domestic satellite issue was one by General Electric that outlines a concept for establishment of a nationwide business-to-business communications system. The selling point for the GE approach is a greatly reduced cost for the transmission of data between business firms or within elements of a widespread organization.

A basic system for use over the U.S., says GE, would require only two satellites; one would be employed in stationary orbit for operational use and the other for backup.

Other plans more specialized

Other proposals for domestic satellites have been filed with the FCC during the last several years, but the majority are for special-purpose relay systems. For example, several of the broadcasting networks presented a joint plan for
Here's the most compact two section variable resistor currently available—the new Allen-Bradley dual Type GD. It's one-half inch in diameter and only a fraction of an inch longer than the popular single section Type G control. The case is dust-tight as well as watertight. Both resistance tracks in the dual Type GD are solid, hot-molded elements, which provide long operating life. As with the single Type G, the noise level is low initially and actually decreases with normal use. Adjustment is smooth at all times with virtually infinite resolution. And low inductance permits operation at frequencies far beyond the usable range of wirewound controls. In addition to standard application, these new dual Type GD controls are ideally suited for use in compact attenuators. Dual Type GD controls are available with nominal resistance values from 100 ohms to 5.0 megohms. For complete specifications on tolerances, tapers, and options, please write Henry G. Rosenkranz, Allen-Bradley Co., 1201 South Second Street, Milwaukee, Wis. 53204. Export Office: 1293 Broad Street, Bloomfield, N.J., U.S.A. 07003. In Canada: Allen-Bradley Canada Limited.
Power is the big problem

Power is the single most important consideration in a satellite communication system. This one factor strongly affects all others. The reason is simple: Generating electrical energy 22,000 miles above the earth is an expensive proposition.

Solar cells are the only practical devices currently available, although nuclear generators have been flown experimentally. The more power a satellite requires, the larger, heavier and more costly will be the array of solar cells that must be orbited to provide it. And high-payload launch vehicles aren't cheap. Intelsat III, with an orbited weight of only 280 pounds, needed a $5.7-million Thor-Delta booster to put it where it is. Since the satellite itself cost only $6-million, the launch cost is a pretty significant fraction of the total.

Another problem with high-power satellites is cooling them. Air and water cooling are a bit difficult in the absence of air and water. All that remains is radiation cooling. This requires large radiators—further adding to the weight problem.

To make the best use of the power available from a satellite (or, putting it another way, to minimize the satellite power needed for a given task) several basic design ideas can be used:

- Extremely large earth station antennas can be employed, together with very low-noise rf amplifiers.
- Directional antennas can be put on the satellite to increase the effective radiated power.
- High modulation-index fm can be used to trade bandwidth for power.
- The satellite's power amplifier can be operated in saturation for better dc-to-rf conversion efficiency.

Unfortunately, each of these ideas has disadvantages as well as advantages, and a trade-off analysis must be made in each case.

Dishing up the power

Perhaps the most straightforward way to increase the information-carrying capability of a satellite system is to use very large antennas and cooled low-noise amplifiers on the ground. The larger the earth-station antennas are made, the more satellite power they will intercept and the less satellite power will be required for each channel.

The economic sense behind this approach can best be illustrated by examining what the International Telecommunications Satellite Consortium (Intelsat) charges the users of its satellites. For a user whose station has an antenna-gain/temperature ratio of 40.7 dB or better, Intelsat charges $20,000 per voice channel per year. A 40.7-dB G/T ratio (G is measured in dB above isotropic, and T is measured in dB above 1°K) requires an earth station with an antenna diameter of approximately 85 feet at the standard down-link frequency of 4 GHz. Such an earth station costs approximately $3-million to $4-million.

Now by dropping down to a 42-foot-diameter dish, one can cut the cost of the earth station to $1-million to $2-million. However, the G/T ratio is now reduced to about 32 dB, requiring about 8.7 dB more satellite power for the same quality of received signal. Because the small-dish earth stations require more satellite power than do the large ones, Intelsat charges them more. The price difference in this case is about $110,000 per voice channel per year—a factor of 6.5! It's obvious that an earth station carrying any reasonable amount of traffic will operate more economically over the long term with the
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Typical printed circuit board used in the MDS 1101 Data-Recorder, showing the extensive use of Allen-Bradley hot-molded ¼ watt resistors.

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A-B hot-molded fixed resistors are available in all standard resistance values and tolerances, plus values above and below standard limits. A-B hot-molded resistors meet or exceed all applicable military specifications including the new Established Reliability Specification at the S level. Shown actual size.
large dish.

It might seem sensible then to use ever larger dishes for more and more economical operation. The rub here is that, at any given frequency, the diameter of the dish must be doubled for every 6-dB increase in antenna gain (Fig. 1). This soon leads to uneconomically large dishes that weigh so much that gravity distorts them from their desired parabolic shape.

85-foot antenna preferred

The 85-foot dish has emerged from trade-off considerations such as these as the more or less standard antenna in international satellite communications. For domestic use, however, some experts feel that a smaller size is best. One reason for this is that domestic earth stations will operate at much higher elevation angles and hence will not need the extremely narrow beamwidths that the big dishes provide to avoid picking up noise from the warm earth. Antenna elevation angles on the order of five degrees are common in international networks when the satellites are placed in orbit above the oceans. But a domestic satellite would, of course, be placed on a longitude line that passes through the U.S., and angles of 30° and more would be typical. This factor will be of great help in preventing interference problems between the satellite's stations on earth and existing terrestrial microwave relay stations.

The same satellite power limitations that make large ground antennas attractive are obviously making television relay directly to the home impractical at this time. Since it is unreasonable to expect every TV set owner to pay for a large dish on his roof, the only way to get adequate signal strength at the receiver is to boost the satellite power. Power levels on the order of tens of kilowatts would be needed for this application. While nuclear power supplies may make this practical in the future, no present planning envisions such powerful satellites.

Instead, TV relay will probably be done between TV studios. A television broadcasting station in a small city might thus incorporate a receive-only earth station on its grounds. Then the TV station could switch its transmitter between its own studio for locally generated programs and the earth station for network programs originating principally in Los Angeles and New York.

It's interesting to note in this application that although the economic considerations are the same basically as in single-pair-of-users application, the resulting optimum system is different. In the single-pair analysis, based largely on existing international services, it was assumed that relatively few earth stations would be using each satellite. It was further assumed that each station would carry a fairly heavy load of traffic (say, a hundred channels). This would justify the construction of expensive earth stations to keep down the cost of the satellite.

A rather different picture may well emerge in the case of a broadcast mode of operation, such as TV relay. Here, a great many earth stations will be available to support the satellite financially, and none will want to spend the millions of dollars needed for the big dish stations. Thus the trend in broadcasting operations will probably be toward smaller earth stations with 15-to-30-foot dishes and toward more powerful satellites.

The extra power can come from increased satellite antenna gain.

Increasing power the easy way

Just as large antennas can be used on the ground to save power, so can they also be used on the satellite itself. There are two major problems to be considered when directional antennas are to be used on a satellite, but both can be solved, as has been proved by Intelsat III and several experimental satellites. The first problem is that of despinning the antenna, and the second is simply that the area of coverage of a satellite and its antenna gain are uniquely related;
To stabilize them in their orbits, satellites are made to spin about an axis parallel to the earth's. The resulting gyroscopic effect maintains their attitude with respect to the earth. It's obvious that for a directional antenna to be kept aimed at the earth, its beam is going to have to move with respect to the satellite. This operation is called despining the antenna.

Despining can be done mechanically, by actually rotating the antenna with respect to the satellite at the same rotational velocity that the satellite has. Or it can be done electronically, by means of a phased array. The disadvantage of the mechanical approach is the decrease in reliability: It's generally a good idea to avoid using moving parts in unmanned spacecraft. This point was made clear when a bearing on an Intelsat III froze up last June. Subsequent models of the satellite had to be modified to avoid the problem, which apparently was caused by the extremely cold temperature of the bearing during the time of the year when it was in the satellite's shadow. The cure was installation of a heater.

Unfortunately, the electronic approach to despining has a worse disadvantage: It wastes power. Phased-array antennas derive their directivity from the interference between the radiating elements in the array. The beam they form is aimed by changing the relative phase of the signal applied to each radiating element (Fig. 2). The most straightforward way to do this is to insert an electrically variable phase shifter between the rf source and each radiating element. Unfortunately, each phase shifter has a certain insertion loss, and the total rf power loss when many elements are used has been deemed too high by the designers of commercial satellites.

Of course, in the future, small, efficient, reliable solid-state rf sources may be available to allow each radiator to have its own oscillator. Then the phasing could be done before the power output stage (Fig. 3), at which point the phaser losses wouldn't be very important. At present, however, traveling-wave tubes provide the best mixture of bandwidth, power, efficiency and reliability, and to have a separate TWT for each radiator seems to be out of the question.

The problems of electrically despun antennas are worse for domestic satellites than for international ones because of their smaller area of coverage. To cover the entire earth, as international satellites do, a beamwidth of approximately 20 degrees is needed. This is generally conceded to require 16 phase shifters at 4 GHz, the frequency normally used for down-link transmission. However, if a 2-degree beam is desired—as it would be in some domestic applications—16 phasers would be needed. The losses would increase with the number of phasers making the mechanical approach even more desirable in a domestic satellite.

The gain-coverage problem

Not much can be done about the relationship between coverage area and satellite antenna gain. The beamwidth of an antenna and its gain are uniquely related (Fig. 4). When one is specified, the other is also. This means that the designer of a satellite really has no freedom of choice between raising the antenna gain or the power level of a satellite once the coverage has been fixed. It also means that the designer of the complete system has a very important choice to make when he decides on the coverage areas.

This is just one of the factors complicating the design of a domestic system. Other technical problems, such as interference reduction, along with political and legal considerations, will ultimately combine with the tradeoff between antenna-gain and satellite power to determine the coverage plan.

Substituting bandwidth for power

An entirely different approach to conserving satellite power, which can be pursued independently of the directional-antenna approach, is to substitute bandwidth for power. This is the same idea that forms the basis of conventional fm broadcasting, but in satellite technology the idea is exploited to the fullest.

A single 4-kHz voice channel is often spread out over 750 kHz in satellite communications by use of extremely high-modulation indices. This can be compared to the conventional fm broadcast situation, where a peak deviation of 75 kHz is used and the baseband signal is 15 kHz wide.

Despite this prodigious use of bandwidth, all commercial satellites in operation today are limited by their total power rather than their bandwiths. (The down-link frequency range for communications
satellites is 500 MHz wide—from 3.7 to 4.2 GHz). Future satellites—for example, Intelsat IV, which is now being developed—will run out of bandwidth before they run out of power if they operate with the same high-performance earth stations that are in use today.

As is well known, the price that one pays for the improvement afforded by an fm system is the existence of a threshold (Fig. 5). This is a value—or more accurately, a range of values—of input carrier-to-noise (C/N) ratio above which the fm receiver operates properly, providing an output signal-to-noise ratio that is considerably higher than the input C/N (the exact amount of the improvement depends on the modulation index and other factors and is thoroughly discussed in many books, such as Panter1). Below the threshold, the system falls apart completely; it's actually worse than a straight a-m system.

Obviously then, it is necessary to operate fm systems above their thresholds. In satellite work where power is so expensive, any technique that can lower the point at which the threshold occurs is naturally going to get a lot of attention. Two of these threshold extension techniques in common use in earth station receivers are the fmfb (fm feedback) and PLL (phaselock loop) approaches.2

In the fmfb approach, frequency feedback is employed to reduce the bandwidth of the i-f signal. A variable-frequency oscillator is made to follow the instantaneous frequency of the incoming modulated carrier. This oscillator serves as the receiver's local oscillator and hence yields a very narrowband i-f signal. The bandwidth of the i-f amplifier may then be reduced, lowering the i-f noise power, and, thus the threshold.

In the PLL approach (Fig. 6) a phase detector replaces the discriminator of the standard fm receiver. The output of the phase detector is low-pass-filtered and used to control the frequency of the reference oscillator of the phase detector. Since the voltage-controlled oscillator is tracking the incoming signal, its control voltage is the output signal. Since the low-pass filter only allows the VCO to track the slowly modulated frequency of the carrier, all of the incoming noise is suppressed except that which falls in a very narrow band around the carrier. Again, this has the effect of lowering the threshold.

The saturation problem

A very obvious way to save power in a power-limited system is to operate all of the high-power devices as efficiently as possible. In the case of satellites, this means operating the output TWT at full power. (The dc power consumption of a TWT is fairly constant over the dynamic range of the tube, so that maximum efficiency occurs at maximum power out.) In addition to providing maximum efficiency, operation of the TWT in a hard-limiting mode has the advantage of making up-link power control unnecessary: The power out of the satellite will be independent of the input power.

As long as a satellite is only going to be used to relay a single carrier, hard limiting is definitely the way to go. Problems arise when you try to use a satellite to relay communications over two separate links at the same time. The highly nonlinear transfer characteristic of the saturated amplifier generates a host of intermodulation products with two deleterious effects: Some of the intermod products fall in the desired signal bands, causing interference, and the others simply eat up a lot of satellite power.

There are essentially two ways out of this dilemma. The first approach, which is in use today, is to back off on the TWT drive power to make the satellite more nearly linear and to continue sending separate messages through the

2. A phased-array antenna aims its beam by changing the relative phase of the signal applied to each radiator. When the phases are all the same, the resultant beam moves to the right, as shown by the blue constant-phase lines. By progressive delay of the phase (from top to bottom), the beam can be tilted up, as shown by the red lines.

3. Phaser losses can be made unimportant through use of a separate power amplifier for each radiator. The amplifiers can be true amplifiers, or perhaps oscillators that are phase-locked to the input signal. In any event, this approach won't be practical until very lightweight reliable solid-state amplifiers are made available.
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4. No freedom exists to specify the antenna gain once the coverage area is chosen. Here antenna gain is plotted against 3-dB beamwidth at a frequency of 4 GHz. The gain equation is given in Fig. 1, the formula for 3-dB beamwidth is \( \theta = 7 \times 10^{4}/fD \), where \( f \) is the frequency in MHz, \( D \) is the antenna diameter in feet, and \( \theta \) is expressed in degrees. (Both formulas are from ITT’s “Reference Data for Radio Engineers,” fourth edition, p 753).

5. The price of fm improvement is the existence of a threshold. In this rather representative system, it occurs in the neighborhood of 10-dB carrier-to-noise ratio. The corresponding S/N is about 25 dB.

The other case of multiple accessing involves a large number of very small users. Internationally these would be the small, emerging nations of Asia and Africa. These users do not have much traffic to send, but they may want to send it almost anywhere. For this application, the Comsat Corp. of Washington, D.C. has developed the SPADE system. This is an FDMA system in which each channel rides on its own carrier and the carriers are assigned on demand. Each station in the SPADE network would be equipped with a frequency synthesizer for the generation of whatever carrier it is told to use when it demands a channel.

The advantage of this approach is its flexibility. No frequency planning is involved since all of the carriers have the same power and bandwidth. This method would probably not be used in a domestic system since terrestrial facilities are adequate for the low-traffic links.

References:
2. Panter, op. cit., Chap. 16 (pps 478-504).
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INFORMATION RETRIEVAL NUMBER 25

EL E CTRON I C D ESIGN 20 , Sept e mber 27 , 1969
Project Sanguine raises fears

Electronic pollution feared

Residents of northern Wisconsin, in growing numbers, are registering concern over the Navy's plans to install a super-power, very-low-frequency communications transmitter system in that region. And it appears that their worries may be justified, since the Navy intends to spend $1.6-million just to determine the effects of such a system on the local environment. Among the worries are these:

What effect will the new system have on other electronics equipment? Will it disrupt telephone services? Will it be dangerous electrically to humans and animals?

Called Project Sanguine, the not-so-secret R&D effort is concerned with a massive transmitting complex that would significantly increase the reliability of the Navy's strategic communications, particularly with its ballistic missile submarine force. Most of the information being released on the project is coming from state and national legislators who are concerned over the environmental effect of the effort.

The Navy reportedly is establishing a Sanguine test facility in Wisconsin that will use a scaled-down design with only a 28-mile crossed antenna grid. Voltages averaging less than 64 volts per meter will be present at the extremities of the grid, according to the Navy, but close to the cable a potential of 100,000 volts per meter may be experienced. The test grid is above ground. The Navy is planning at least a year of testing.

Should the system ultimately be approved, it has been estimated that it will require 6000 miles of buried cable and 240 transmitter sites each requiring 10 acres. The buried cable will require 30-foot rights-of-way between the fenced-off transmitter sites. Neither the operating frequency nor power output has yet been revealed.

With the multi-megawatt power believed contemplated in this system, opponents are afraid the Navy station may induce currents in every metallic object in or near the transmitter area. The Navy has admitted transmissions could induce "mild electric shock," but has indicated it plans to insulate at regular intervals everything in the area that needs to be insulated.

A principal opponent of the new system is Democratic Senator Gaylord Nelson. He says Navy figures show that 1000 feet of wire (such as in a farm fence) 100 feet from the underground antenna will carry a potential of 52 volts, and that the same wire at two miles will carry 22 volts. He also notes that a typical farmer's electrified fence carries only 12 volts and that under certain conditions even this has been known to be dangerous.

A principal Navy contractor investigating the effects of the planned facility is Hazelton Laboratories of Falls Church, Va. It has a $150,000 contract to study such effects. But, based on its original proposal, some critics feel that Hazelton's ultimate report may be somewhat biased toward Navy wishes.

In its proposal, the firm noted that because of the intensity of the transmitter output, many local animals and plants would be exposed to the "field for very long periods of time, unless they are killed by it, are removed or, in the case of free-ranging animals, are driven out by noxious effects." The firm also pointed out a potential hazard to the local ecology resulting from increased temperature effects during transmitter operation. Yet it proposed only to make a limited study of this effect, using plants. It stated further "that such animals as groundhogs are not of sufficient economic value to
be of concern. Hazelton does plan radiation tests on a variety of domestic animals in its laboratory here.

Other contracts issued by the Navy in the test program included the following:
- Bell Telephone Laboratories, $200,000 to study station influences on area telephone services.
- Illinois Technical Research Institute, $400,000 to study ways to minimize interference with other electronic equipment.
- RCA, $500,000, to look into electrical effects on nearby fencing and railroads.

Military communications study due
The Pentagon's Defense Communications Agency, which directs the strategic military global communications network, has established a study group to investigate problems in linking its system to the tactical communications systems contemplated by the U.S., Australia, Canada and Britain. There is especial concern over the equipment that will be developed in the current four-nation R&D program called Project Mallard.

Mallard was begun in 1965 and is intended to produce by the mid-1970s the designs and basic hardware for a highly advanced digital tactical system for common use by the field forces of the four countries. The system will involve all communications modes, including satellite relay, and will employ automatic switching throughout.

Inertial sensor improved
Northrop Corp. representatives here are eloquent in their praise of an advanced inertial sensor recently developed by the company's Electronics Div. in Norwood, Mass. The unit combines in a single instrument a two-axis gyro and a two-axis accelerometer. Thus, they report, two of these elements can be combined by installing them perpendicularly to one another to create a complete inertial reference, a three-axis system plus one redundant axis. Called a Multi-Function Sensor, the miniaturized prototype element is 2.1 inches long by 0.7 inch diameter and weighs just under 125 grams. The unit has been under development for seven years, the company says. Specifications for the unit indicate a sensitivity acceptable for an aircraft application, possibly in a strap-down (gimballess) inertial navigation system.

Navy gets new computer controls
The Navy Supply Systems Command here reports that the first of seven new inventory-control computer systems has begun operation at the Jacksonville Naval Air Station in Florida. Six more air stations will be equipped similarly during the next year with the paired IBM System 360 computers, one a Model 50 and the other a Model 25. When the full system is complete, it will control parts for and support over 50 aircraft types in use by the Navy.

Under its contract awarded last June, IBM provides the basic computers for each facility plus all the remote terminals and interconnecting communications equipment to tie in with the central computers. Also, each computer complex employs a magnetic disc storage unit with a 233-million character capacity.

The Navy says that the computers will provide "inventory control of one million aircraft parts valued at $2 billion." The seven repair centers annually process five million requisitions and two million receipts.

'Brain-drain'—fact or fiction?
A report published by the Population Reference Bureau in Washington, examines the so-called "brain-drain"—the influx of scientists and other highly skilled people to this country from foreign lands. A total of 12 per cent of all recent immigration to the United States consisted of such highly skilled people, the report notes. It says that recent immigration to the U.S. has ranged from 306,000 to 360,000 annually. Of this total 8682 were engineers in 1967.
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<table>
<thead>
<tr>
<th>Series</th>
<th>Contact Combinations</th>
<th>Center-to-Center Pin Spacing</th>
<th>Poles</th>
<th>Construction</th>
<th>Special Features</th>
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<tr>
<td>812 Miniature</td>
<td>Form A, Form B, Form C (A + B)</td>
<td>.150 Inch</td>
<td>Up to 6</td>
<td>Open Type</td>
<td>Economy miniature series</td>
</tr>
<tr>
<td>822 Miniature</td>
<td>Form A, Form B, Form C (A + B)</td>
<td>.100 Inch</td>
<td>Up to 4</td>
<td>Open Type</td>
<td>Smallest package using miniature reed switch</td>
</tr>
<tr>
<td>823 Miniature</td>
<td>Form A, Form B, Form C (A + B)</td>
<td>.100 Inch</td>
<td>Up to 4</td>
<td>Potted in Plated Metal Can</td>
<td>Environmental protection</td>
</tr>
<tr>
<td>824 Miniature</td>
<td>Form A, Form B, Form C (A + B)</td>
<td>.100 Inch staggered</td>
<td>Up to 6</td>
<td>Epoxy Cased, Potted in Pliable Compound</td>
<td>High reliability version</td>
</tr>
<tr>
<td>842 Standard</td>
<td>Form A, Form B, Form C (A + B) and True Form C</td>
<td>.200 or .218</td>
<td>Up to 8</td>
<td>Open Type</td>
<td>Higher power control</td>
</tr>
<tr>
<td>843 Standard</td>
<td>Form A, Form B, Form C (A + B) and True Form C</td>
<td>.200 or .218</td>
<td>Up to 8</td>
<td>Potted in Plated Metal Can</td>
<td>Environmental protection</td>
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<tr>
<td>852 Standard</td>
<td>One Form A only</td>
<td>.250 Inch staggered</td>
<td>1 pole only</td>
<td>Open Type</td>
<td>Economy standard relay</td>
</tr>
</tbody>
</table>
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SIDELIGHTS

Are we overcrowding space?

The last question that concerned ELECTRONIC DESIGN staffers Charlie LaFond and Mike Riezenman while they were writing their stories on prospects for a U. S. domestic satellite program, (pp. 36 to 48), was the crowding of space with flying objects. Charlie and Mike were largely occupied with lessons of the recent past and problems of the immediate future. But several people they interviewed—notably Wilbur L. Pritchard, director of the Comsat Laboratories in Washington, D. C.—were very much concerned with the eventual effect of placing so many satellites in orbit.

At first thought, this problem may seem too remote to cause worry. After all, at the synchronous altitude of approximately 22,000 miles, more than 150,000 miles of orbital space are available for satellites.

But the hard facts are that there are certain preferred orbital positions, and certain minimum separations must be maintained between satellites. If the “birds” get much closer to each other than about two degrees (as seen from the earth) then earth stations will have a tough time trying to get their messages one at a time without interference.

The time to solve this problem, says Pritchard, is now—not after the sky is already crowded and various countries and companies are feverishly competing for the little remaining space.

And just in case anyone still thinks that it’s too early to worry about space crowding, Pritchard points out that only 60 years ago, when Bleriot first flew the English Channel, people would have thought you crazy if you’d started planning to avoid major air-traffic tieups over New York City.
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INFORMATION RETRIEVAL NUMBER 30

ELECTRONIC DESIGN 20, September 27, 1969
EDITORIAL

Launch—or get off the pad

Just about everyone agrees that the United States had better get moving on the establishment of a domestic communication satellite system. Unfortunately, there is something less than full agreement on the form the system should take, the material it should carry and, especially, who should ultimately own and operate it.

Comsat believes it has a Congressional mandate to own and operate such a system. AT&T says no—it is all well and good for Comsat to own and operate the satellite portion of the system, but the ground stations should be the properties of the common carriers. And General Electric, which has also proposed a system, says it doesn't care who owns or operates it.

To make matters worse, there is some speculation that the Federal Communications Commission, which to date has refused to reach a decision without knowing the Administration's position, may not even have the legal power to decide the issue—and that a final solution may have to come from the courts, after the FCC reaches its decision.

The end result is that years will have elapsed between the first serious proposals of Comsat, AT&T and the Ford Foundation and the actual launching of the first satellite in a domestic system.

During this time the rest of the world will not be standing still. As pointed out in the Special Report beginning on page 36 of this issue, four countries as of today are either actively considering or have in operation domestic communication satellite systems. And who knows what other countries will join the ranks within the next two years?

If the FCC commissioners do not soon get their decision off the launching pad, the United States, with its great expertise in communications satellites, may one day find itself akin to the cobbler's children who had no shoes.

We feel the time has come for the FCC to stop procrastinating and waiting for the Administration to make its position known.

A decision should be reached as soon as possible—one based on the present and future communications requirements of the nation. Any changes in that decision can then be left to the courts and/or the Congress.

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- Rugged mechanical stops for dependability.
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**Brief Specs**

<table>
<thead>
<tr>
<th>Size:</th>
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<tbody>
<tr>
<td>Resistance Range:</td>
<td>15 ohm to 180K</td>
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<tr>
<td>Resistance Tolerance:</td>
<td>±5%</td>
</tr>
<tr>
<td>Independent Linearity:</td>
<td>±0.25%</td>
</tr>
<tr>
<td>Power Rating:</td>
<td>3 watts @ 40°C</td>
</tr>
</tbody>
</table>

The model 532 is available through your local Spectrol distributor. For full specs, circle the reader service number. Qualified respondents may obtain a sample free of charge through their Spectrol representative.

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**Brief Specs**

<table>
<thead>
<tr>
<th>C.P.</th>
<th>Cermet</th>
</tr>
</thead>
<tbody>
<tr>
<td>Model 138</td>
<td>Model 139</td>
</tr>
</tbody>
</table>

- **Size:** 1-5/16", 1-5/16"  
- **Resistance:** 5000 to 1 Megohm  
- **Range:** 100KΩ  
- **Independent Linearity:** ±0.5%  
- **Power Rating:** 2 w @ 40°C, 5 w @ 40°C

With either choice you get the same "designed-in" reliability and rugged construction as with our Model 132 wirewound—and all are priced lower than the competition! For full specs, circle the reader service number. Qualified respondents may obtain a sample free of charge through their Spectrol representative.
General models of nonlinear devices are proving valuable in computer-aided design. Oscilloscope and X-Y plotter curves are key elements in determining a model. p. 68.

Precision management, with bar charts and other aids to keep pace of progress, helped Grumman engineers assemble and test the LM 5 for its landing on the moon. p. 78.

Also in this section:

Streamline feedback-amplifier design with a simple analysis. p. 74.

Ideas for Design. p. 84.
Develop useful general models
of nonlinear devices for computer-aided design.
The tunnel diode illustrates the curve-fitting approach.

Computer-aided circuit analysis requires a mathematical model for each circuit element. In particular, nonlinear devices must be accurately modeled. Transistor and switching diode models have been well developed and documented, but SCRs, UJTs, and tunnel diodes, among others, usually have to be devised by the designers themselves.

Models may be classified by range of validity and complexity. Ideally, a model should be frequency independent, valid for any current or voltage excitation within the device ratings, and have a fixed topology that provides a one-to-one correspondence, with physical processes. It should consist of passive elements (R, L, C and mutual inductance), if possible of fixed value, current sources (I) and voltage sources (V) whose mathematical equations are "well-behaved."

Small-signal, piecewise linear and large-signal models are all nonlinear simulations. A large-signal model, because it can be used over a wide range of operation, is the type best suited for use in general computer circuit analysis. It may be more complex than necessary for many applications, but it is usually safer to simplify a general model than elaborate a simple one. For these reasons the model discussed in this article is of the large-signal variety.

Formulate general device models

But whatever family of semiconductor devices is involved, certain general considerations and requirements must be regarded:

- An understanding of the physical processes that produce the external electrical characteristics is desirable. Although it is possible to use a purely black-box approach, it is helpful to know which phenomena are the result of tunneling, avalanche of the junction, stored charge, conductivity modulation, etc.
- The manufacturing process used in making a given device may be pertinent to the model. For example, a transistor of planar construction will probably exhibit different transient behavior from a transistor having the same 2N number although of annular construction. The result may need to be two different models or two different sets of model parameters, one for each manufacturer or manufacturing process.
- A major portion of any model is the set of mathematical equations that describes the physical operation. They may be derived in various ways, but two criteria that must be met are an adequate description of device behavior and compatibility with the language of the computer.
- The model topology is the interconnection of the discrete components making up the network that approximates the device. Although not imperative, it is usually desirable to have one element or set of elements to represent each physical phenomenon. In a tunnel diode, for example, one resistor is used for series resistance, and one capacitor represents junction capacitance; but one current source simulates tunnel current and a second represents minority carrier current.
- The development of a model involves a determination of certain defining parameters. These parameters must be obtained from knowledge of the device physics or from operating characteristics. If the operating characteristics are not supplied by the manufacturer, then measurements must be made by the user. Sometimes it is more difficult to determine what measurements to make and how to make them than it is to formulate the equations and the topology.
- Once the model has been formulated, the device parameters must be obtained. Usually it is necessary to extract these parameters from the data measurements, previously mentioned. The data reduction involves curve-fitting data points or other analytical manipulations to get the parameters explicitly.

An important constraint on the mathematical equations is that they must have the same form as the network equations used in the general program. For example, in finding a steady-state solution using the SCEPTRE program, a current source cannot be a function of the voltage across
it except for the special case of the diode equation, \( I = I_0 (e^{V_i} - 1) \). This constraint only means that, when putting a model into a general program, one should be aware of any unique features of that program.

**Tunnel-diode model is examined**

A general approach to the modeling of any nonlinear device is the technique of curve fitting. A computer can take the tedium out of the job. A tunnel-diode model illustrates the curve-fitting technique. The voltage-current characteristic curve of a typical tunnel diode is shown in Fig. 1 along with some of the important dc parameters.

The unique properties of a tunnel diode that are most often utilized are its negative resistance and its very rapid switching from \( V_p \) to a point in the forward conduction region (at a voltage something less than \( V_{fp} \)). For voltages greater than \( V_{fp} \), the tunnel diode follows the ordinary diode characteristic.²

Most tunnel-diode circuits are designed around the peak point and valley point, so any tunnel-diode model should accurately approximate these points. It is relatively easy to measure the peak point, valley point, and the maximum slope in the negative conductance region (\(-G\)). Measurement of current at voltages greater than \( V_{fp} \) is also quite easily accomplished with automatic test equipment.

There is very little resistance to current flow when reverse bias is applied to the junction — that is, with reverse bias, the tunnel diode is nearly a short circuit. Although the model to be developed is valid for negative bias, accurate modeling of reverse behavior is normally considered of little importance in tunnel-diode applications.

**Establish model topology**

The topology assumed to represent a tunnel diode with lumped components is shown in Fig. 2. The values of the series resistance, \( R_s \), and the series inductance, \( L_s \), are normally given on manufacturers' specification sheets. The series inductance is usually of the order of 1 to 10 nanohenries (depending on the length of the leads) and must be included because of the high normal operating frequencies.

The junction capacitance, \( C_j \), can be measured on a double-balance rf admittance bridge.

Although no capacitance measurements can be made in the negative resistance region, sufficient data points are available for \( V < V_p \) and \( V > V_v \) to determine that a typical junction capacitance versus \( V \) curve has the general shape shown in Fig. 3. The \( C_j \) curve is described by the same relationship used for ordinary diode transition capacitance; namely,
The constants $K$, $V_z$, and $N$ must be evaluated from the plot of $C_i$ vs. $v$. Typical values are $N$ between 0.3 and 0.5 and $V_z$ between 0.6 and 1.0. $K$ varies more widely and is proportional to $C_i$ since it is dependent on junction geometry.

The current generator, $I$, approximates the static V-I characteristic shown in Fig. 1. Development of a mathematical model includes finding an expression for current as a function of junction voltage, $v$. It must be possible to obtain any unknown coefficients in the defining equations from external measurements on the device, or from published data. One effective method of obtaining values for the peak and valley point, and points in the forward conduction region is to photograph the face of an oscilloscope curve tracer. Measurements are then made from the photographs. Figure 4 shows several examples of this technique.

### Write curve-fitting equations

Several different methods for modeling the current generator $I$ have been developed. One approximates the entire characteristic by an exponential function

$$I(v) = A v e^{-a v} + B (e^{b v} - e^{-b v}) + C (e^{c v} - 1)$$

(1)

Where $I(v)$ is the current through the diode as a function of the voltage ($v$) across it and $A$, $a$, $B$, $b_1$, $b_2$, $C$ and $c$ are constants to be determined.

Equation 1 represents the sum of the tunneling current, the excess current, and the diffusion current, respectively; or $I(v) = I_{\text{tu}} + I_{\text{ex}} + I_{\text{diff}}$.

The entire V-I characteristic might be fitted exactly to Eq. 1, but this is not always feasible. First, the entire curve would have to be known; second, the solution of one equation in seven unknowns can be tedious at best. Without approximations and assumptions to reduce the number of unknowns, a computer search routine to do the job is relatively inefficient.

It seems, however, that with what is known about the unique tunnel-diode curve and some assumptions, a reasonably good approximation can be made without knowing the entire characteristic. A procedure for obtaining the equation coefficients with only a limited amount of information will now be developed.

First, measure four points: the peak point, $[V_p, I_p]$, the valley point, $[V_v, I_v]$, and two points in the forward conduction region, $[V_m, I_m]$ and $[V_1, I_1]$, where $V_m > V_1 > V_p$. The last two points are normally chosen so that $I_m$ is near the maximum rated current of the device and $I_1$ is about 1.5 $I_p$. The value of $-G$ is also needed. It can either be taken from most diode specification sheets or it can be measured.

For $v > V_{fp}$, the last term of equation (1) is dominant, so it can be assumed that

$$I(v) = C (e^{cv} - 1), v > V_{fp}.$$  

Using the two data points in the forward conduction region, $[V_m, I_m]$ and $[V_1, I_1]$, and simplifying, the results are:

$$c = \frac{\ln(I_m)}{V_m - V_1}$$

and

$$C = \frac{I_f}{(e^{v_f} - 1)} = \frac{I_m}{(e^{v_m} - 1)}$$

4. Oscilloscope traces of tunnel-diode characteristics facilitate measurement of peak, and valley points. The loss of detail in the negative resistance region of the 1N2934A makes measurement of $V_1$ difficult.
Thus \( c \) and \( C \) are known.

Since the peak point is a relative maximum and the valley point is a relative minimum, the first derivative of \( I(v) \) must be zero at these points. For \( v < V_p \), it can be shown that the diffusion current is negligible \([C(e^{cv} - 1) \approx 0]\). When \( v \approx V_p \), none of the three current expressions is negligible, although \( I_{ex} \), the excess current, is the majority term. Therefore, from Eq. 1 and its first derivative evaluated at \( (V_p, I_p) \) and \( (V_v, I_v) \) respectively:

\[
I_p = AV_p e^{-b_1 V_p} + B(e^{b_1 V_p} - e^{-b_2 V_p}) \quad (2)
\]

\[
I'(V_p) = O = A e^{-a V_p} (1 - a V_p) + B(b_1 e^{b_1 V_p} + b_2 e^{-b_2 V_p}) \quad (3)
\]

\[
I_v = AV_v e^{-b_1 V_v} + B(e^{b_1 V_v} - e^{-b_2 V_v}) + C(e^{cv} - 1) \quad (4)
\]

\[
I'(V_v) = O = A e^{-a V_v} (1 - a V_v) + B(b_1 e^{b_1 V_v} + b_2 e^{-b_2 V_v}) + cC e^{cv} \quad (5)
\]

This gives four equations in five unknowns. One more item of information is known; the maximum negative conductance. Since \(-G\) is the maximum slope of the negative conductance region, it is at the point of inflection, \( V_i \), of \( I(v) \). However, since \( V_i \) is not known, some approximations must be made. The diffusion current has very little effect, and the excess current is nearly linear in this region. Therefore, assuming that the inflection point of \( A e^{cv} \) is also the inflection point of \( I(v) \), gives \( V_i = \frac{2}{a} \). Then,

\[
I'(V_i) = -G = A e^{-a V_i} (1 - a V_i) + B(b_1 e^{b_1 V_i} + b_2 e^{-b_2 V_i}). \quad (6)
\]

Equations 2 through 6 constitute a set of equations that must be solved to obtain \( A, a, B, b_1 \), and \( b_2 \). A method of iteration which can easily be programmed can be used to obtain the solution as follows:

Assume initial values for \( B, b_1, b_2 \). Using these values, find \( A \) and \( a \) from Eqs. 2 and 3. Substitute these calculated values of \( A \) and \( a \) into Eqs. 4, 5 and 6. Eliminate \( B \) from this set of equations, and there remain two transcendental equations in \( b_1 \) and \( b_2 \).

\[
b_1 e^{b_1 V_i} + b_2 e^{-b_2 V_i} = K_1 \quad (7)
\]

\[
b_1 e^{b_1 V_i} + b_2 e^{-b_2 V_i} = K_2 \quad (8)
\]

The last two equations provide a definition of \( b_1 \) and \( b_2 \). Several iteration procedures are available for solving such equations. Perhaps the simplest method is the linear iteration. (See Reference 6 for a discussion of other methods that might be used.)

### Linear-iteration method is simplest

To apply the linear-iteration method, begin with Eqs. 7 and 8 rearrange them into a form

\[
b_1^{(j + 1)} = g_1(b_1^{(j)}, b_2^{(j)}) \quad (9)
\]

\[
b_2^{(j + 1)} = g_2(b_1^{(j)}, b_2^{(j)}) \quad (10)
\]

The iteration described by Eqs. (9) and (10) is

---

**Flow chart of computer program** shows the steps to be followed in determining parameters of the curve-fitting equation. Note the iteration steps for the \( a, A, b \) and \( B \) parameters.
6. Static V-I characteristic for 1N2929 tunnel diode illustrates the closeness to fit that can be obtained by the curve-fitting method.

continued until \( b_1 \) and \( b_2 \) differ from \( b_1 \) and \( b_2 \) by less than some preassigned amount.

After \( b_1 \) and \( b_2 \) have been calculated, a new value of \( B \) is found from Eq. 4.

\[
B = \frac{K_3}{e^{b_1 V} - e^{-b_2 V}}.
\]

The new \( B, b_1, \) and \( b_2 \) found are substituted into Eqs. 2 and 3 to calculate better values for \( A \) and \( a \).

The entire iteration is continued until the parameter values converge within a predetermined accuracy. A flow chart of the complete iteration is shown in Fig. 5. Results from this method are presented in Fig. 6.

The advantages for this method of describing \( I(v) \) are: (a) the need for only specification-sheet information, plus two easy measurements in the forward region; and (b) the fit is exact at the two most critical points — peak point and valley point. The disadvantages are that a good fit to the entire characteristic is not always possible, and that the five equations relating \( A, a, B, b_1, \) and \( b_2 \) are difficult to solve; the iteration sometimes doesn't converge.

Another method of curve fitting is to divide the V-I characteristic into zones and fit an \( n \)th-degree polynomial function to each zone. The degree of fit can be made as good as desired by using more zones and a larger \( n \).

For the tunnel-diode example, three zones, one below the peak point, one including the peak point and one including the valley point and the forward conduction region, and up to 5th-degree polynomials give a fit to within 3%. The constraints on the polynomials are: all functions and their first derivatives must be continuous, and functions for adjacent zones must have matching values and first derivatives at the zone boundary.

The advantages of this method are that polynomial functions are well-behaved and easy to manipulate, the polynomial coefficients are easy to extract, and a very close fit can be obtained. The disadvantages are that each polynomial has a limited zone of applicability, and the entire device V-I characteristic must be accurately known.

The key to the successful use of these methods is to use the computer to eliminate the mathematical drudgery. Thus, the computer that created the need for accurate device models is the means by which the models themselves are generated.

References


Test your retention

Here are questions based on the main points of this article. Their purpose is to help you make sure you have not overlooked any important ideas. You'll find the answers in the article.

1. Which type of nonlinear device model is most general?

2. What nonmathematical constraints must be placed on the model equations?

3. How does one obtain the data from which to construct a tunnel-diode model?

4. What three significant currents make up the total tunnel-diode current?
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Streamline feedback-amplifier design. It’s simple if you analyze forward and reverse transmission paths of your circuit separately.

In designing feedback amplifiers the forward and reverse transmission paths must be separated. The brute-force approach is to write loop equations and manipulate them into a form that permits isolation of the two paths. This is time-consuming, however, and tends to obscure the dominant parameters. A simpler method is to analyze the paths separately, taking into account the loading effects of the source, load and feedback impedances.

Actually, the impedance levels of the various feedback methods indicate that a different set of equations is best suited to each case. Thus, selection of the proper matrix not only reduces design time but also offers better insight into circuit performance.

Consider the general case

The general feedback amplifier shown in Fig. 1a has the transfer function:

\[ G = \frac{\alpha}{1-\alpha \beta} \]  

(1)

where \( \alpha \) is the forward amplifier gain and \( \beta \) is the feedback function.

Two assumptions are made in the analysis of such an amplifier:

- First, the forward transmission through the feedback path is negligible in comparison with the transmission through the forward amplifier path;
- Second, the reverse transmission through the forward amplifier path is negligible in comparison with the reverse transmission through the feedback path.

Figures 1b and 1c consider the combination of the closed-loop amplifier and the source impedance as a two-port network. The input parameters are the source voltage \( (E_s) \) and the source current \( (I_s) \). The output parameters are the load voltage \( (E_o) \) and the load current \( (I_o) \). \( I_o \) is set equal to minus \( I_s \) since the two-port analysis defines currents flowing into the positive terminal.

There are four general types of interconnections between the forward and feedback paths to be considered:

- Series-shunt feedback, which provides a high input impedance and a low output impedance.
- Shunt-series feedback, which provides a low input impedance and a high output impedance.
- Shunt-shunt feedback, which provides a low input and output impedance.
- Series-series feedback, which provides a high input and output impedance.

Series-shunt feedback

The series-shunt feedback amplifier shown in Fig. 2 lends itself to an \( h \) parameter two-port. In matrix form, its equations are:

\[
\begin{bmatrix}
E_s \\
-I_o
\end{bmatrix}
= 
\begin{bmatrix}
h_{11} + Z_s & h_{12} \\
h_{21} & h_{22}
\end{bmatrix}
\times
\begin{bmatrix}
I_s
\end{bmatrix}
\]  

(2)
Solving for voltage gain:

\[
\begin{align*}
\frac{E_o}{E_i} &= \frac{-h_{21}}{(h_{11}+Z_s)} \left( h_{22} + Y_L \right) \\
&= \frac{1}{1 - h_{21}h_{11}} \left( h_{11} + Z_s \right) \left( h_{22} + Y_L \right)
\end{align*}
\]  

(3)

Comparing Eq. 3 with the general form (Eq. 1) yields:

\[
\alpha = \frac{-h_{21}}{(h_{11}+Z_s)} \left( h_{22} + Y_L \right)
\]

(4)

\[
\beta = h_{21}
\]

Using standard two-port technique, the terms in Eq. 4 can be related to circuit values. The results are summarized in Table 1.

Shunt-series feedback

The shunt-series feedback amplifier shown in Fig. 3 lends itself to a \( g \) parameter two-port, whose equations in matrix form are:

\[
\begin{bmatrix}
I_s \\
E_s
\end{bmatrix} = \begin{bmatrix}
g_{11} + Y_s & g_{21} \\
g_{21} & g_{22}
\end{bmatrix} \times \begin{bmatrix}
E_o \\
I_o
\end{bmatrix}
\]

(5)

Solving for current gain:

\[
I_o = g_{21} \frac{E_o}{(g_{11}+Y_s) (g_{22}+Z_L)}
\]

(6)

\[
I_s = \frac{1 - g_{11}g_{21}}{(g_{11}+Y_s) (g_{22}+Z_L)}
\]

Comparing Eq. 6 with Eq. 1 yields:

\[
\alpha = \frac{g_{21}}{(g_{11}+Y_s) (g_{22}+Z_L)}
\]

(7)

\[
\beta = g_{11}
\]

The results of these calculations are summarized in Table 1.

Shunt-shunt feedback

The shunt-shunt feedback amplifier shown in Fig. 4 lends itself to the use of a \( y \) parameter two-port:

\[
\begin{bmatrix}
I_s \\
E_s
\end{bmatrix} = \begin{bmatrix}
y_{11} + Y_s & y_{12} \\
y_{21} & y_{22}
\end{bmatrix} \times \begin{bmatrix}
E_o \\
I_o
\end{bmatrix}
\]

(8)

Solving for current gain:

\[
I_o = -y_{21} \frac{E_o}{(y_{11}+Y_s) (y_{22}+Y_L)}
\]

(9)

\[
I_s = \frac{1 - y_{11}y_{21}}{(y_{11}+Y_s) (y_{22}+Y_L)}
\]

2. Typical series-shunt configuration (a) is best analyzed using \( h \) parameters (b) since \( I_s \) is common to both the forward and reverse networks and \( E_o \) is across both networks.

3. Typical shunt-series configuration (a) is best analyzed using \( g \) parameters (b) since \( I_o \) is common to both the forward and reverse networks and \( E_o \) is across both networks.

4. Typical shunt-shunt configuration (a) is best analyzed using \( y \) parameters (b). This selection of parameters is based on the fact that \( E_o \) and \( E_s \) are both across the forward and reverse networks.
Table 1. Design equations

<table>
<thead>
<tr>
<th>Configuration</th>
<th>Equation</th>
<th>$\beta$</th>
<th>$\alpha$</th>
</tr>
</thead>
<tbody>
<tr>
<td>Series-shunt</td>
<td>$I_0 = \frac{R_{E_1}}{R_{E_2} + R_f}$</td>
<td>$E_o = \frac{1}{R_{E_1} + R_{E_2} + R_f} \left( \frac{R_{L_1}}{R_{L_1} + R_{E_2} + R_f} \right) \left( \frac{R_{L_2}}{R_{L_2} + R_{E_2} + R_f} \right)$</td>
<td>\begin{align} a &amp;= \frac{y_{12}}{y_{21}} \left( y_{11} + y_{12} \right) \ \beta &amp;= -\frac{y_{12}}{y_{11}} \end{align}</td>
</tr>
<tr>
<td>Shunt-series</td>
<td>$I_L = \frac{R_p}{R_{E_1} + R_{E_2} + R_f}$</td>
<td>$E_o = \frac{1}{R_{E_1} + R_{E_2} + R_f} \left( \frac{R_{L_1}}{R_{L_1} + R_{E_2} + R_f} \right) \left( \frac{R_{L_2}}{R_{L_2} + R_{E_2} + R_f} \right)$</td>
<td>\begin{align} a &amp;= \frac{y_{12}Z_1}{y_{12}Z_2} \ \beta &amp;= \frac{y_{12}}{y_{12}Z_1} \end{align}</td>
</tr>
<tr>
<td>Shunt-shunt</td>
<td>$R_L = R_f$</td>
<td>$E_o = \frac{R_f}{R_{E_1} + R_{E_2} + R_f}$</td>
<td>\begin{align} a &amp;= \frac{y_{12}}{y_{12}Z_1} \ \beta &amp;= \frac{y_{12}}{y_{12}Z_1} \end{align}</td>
</tr>
<tr>
<td>Series-series</td>
<td>$R_f = R_L$</td>
<td>$E_o = \frac{R_f}{R_{E_1} + R_{E_2} + R_f}$</td>
<td>\begin{align} a &amp;= \frac{y_{12}}{y_{12}Z_1} \ \beta &amp;= \frac{y_{12}}{y_{12}Z_1} \end{align}</td>
</tr>
</tbody>
</table>

Comparing Eq. 9 with Eq. 1 yields:
\begin{align} a &= \frac{-y_{21}Y_L}{(y_{12} + Y_L)(y_{11} + Y_E)} \\ \beta &= -\frac{y_{12}}{y_{11}} \end{align}

The results of these calculations are summarized in Table 1.

**Series-series feedback**

The series-series feedback amplifier shown in Fig. 5 lends itself to the use of a $z$ parameter two-port:
\begin{align} \left[ \begin{array}{c} E_o \\ E_{o2} \end{array} \right] &= \left[ \begin{array}{cc} z_{11} + Z_s & z_{12} \\ z_{21} & z_{22} \end{array} \right] \times \left[ \begin{array}{c} I_s \\ I_{o2} \end{array} \right] \\ & \quad \text{(11)} \end{align}

Solving for voltage gain:
\begin{align} E_o &= \frac{z_{21}Z_s}{(z_{11} + Z_s)(z_{22} + Z_s)} E_{o2} \\ E_{o2} &= \frac{1 - z_{21}z_{21}/(z_{11} + Z_s)(z_{22} + Z_s)}{z_{21}Z_s/(z_{11} + Z_s)(z_{22} + Z_s)} E_o \\ & \quad \text{(12)} \end{align}

Comparing Eq. 12 with Eq. 1 yields.
\begin{align} a &= \frac{z_{12}Z_s}{(z_{11} + Z_s)(z_{22} + Z_s)} \\ \beta &= \frac{z_{12}Z_s}{(z_{11} + Z_s)(z_{22} + Z_s)} \end{align}

The results of these calculations are summarized in Table 1.

**Reference**

For 60¢ you can buy a diode with 100 picosecond switching time, low forward resistance, low reverse current, low stored charge and all the other parameters that add up to better-than-PN junction performance. Hewlett-Packard's 5082-2810/11 Hot Carrier Diodes are the best—coming and going—that you can buy for high-speed switching at low cost. They're 4½ times faster than any PN junction. Try to beat that combination in a circuit design.

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Richard L. Turmail, Management & Careers Editor.

Immediately before Apollo 11’s module scored a lunar touchdown, most of us back on earth were huddled around TV screens nervously anticipating all the mechanical and electrical functions that could go wrong. The happy fact that nothing did go wrong is a tribute to NASA’s Lunar Module program in general, and to those responsible for the success of the LM assembly and test program in particular.

Because we think our readers are interested in knowing exactly how such a program was managed, we interviewed Paul Butler, Grumman Aerospace Corporation’s Assistant LM Program Director for SCAT (spacecraft assembly and test). Mr. Butler is currently working on the company’s space station project. NASA awarded the LM contract to Grumman in January, 1963. The company accounted for about 45% of the cost of the project, with over 7000 suppliers accounting for the remaining percentage.

Our questions and Butler’s comments follow:

What were your emotions when the LM was approaching a lunar landing?

The landing didn’t worry us, especially when Armstrong said that systems were still “go.” The alarm light that came on didn’t bother us because we knew the astronauts had more than 30 seconds of fuel left.

What worried us most was the launch from the moon’s surface. During the landing the astronauts had an abort capability at any given time. Once they were on the moon, however, the engines had to work the first time. It bothered us that a man couldn’t take over until the LM engine lit. I guess all of us were mentally going through the things that had to function to allow the ascent stage to leave the moon.

How is the LM project organized?

Dr. Ralph Tripp, LM Program Director, reports to Joe Gavin, Vice President, Space Programs. Five assistant program directors serve him through (1) engineering; (2) manufacturing; (3) spacecraft assembly and test; (4) ground support equipment; and (5) quality control.

What was SCAT’s responsibility?

SCAT’s responsibility was to assemble and test the LM to delivery for launch. Manufacturing produced the basic structural shell and wiring harness of the LM, and our job was to install all the black boxes, outfit the cabin, test all the systems, and take it up to the point where it could be flown. When the vehicle left Grumman it was absolutely complete, except for the fuel and the computer programs that were designed for the moon mission.

How did you maintain quality control?

Grumman representatives were present at each vendor’s plant. We changed the mix of engineering and quality control constantly; we let it pulse as a function of what we were after. The vendors checked their parts out of their plant, and we checked them into our plant. We ran what we called a PIT-test (pre-installation test) every time we got a new piece of equipment. We’d check every piece before installing it in a vehicle to make sure it had not been damaged in transit.

Every time we broke, say, a connector, before we placed it back in its box we had to go back into the circuitry and exercise every copper path that went through that plug to be sure that the pin mated, and to be sure that when it didn’t mate it didn’t bend or come loose. Every single wire had to be functioning and have continuity from end to end.

How did you coordinate the work?

Coordination was the most effective key to the entire program. I happen to be a nut on planning and scheduling. Planning, to me, isn’t tracking. Tracking just tells me where I am today, and I never want to know just that. What I want to
“You pay a guy a salary to do a professional job. If he does a good job, why do you have to reward him?”

“In development, all I can really guarantee is that I will never make the same mistake twice.”
The LM engine ignites! Grumman engineers congratulate each other as the lunar module kicks off the moon.

know is—based on where I am today—what's left, and how do I get there from here. So I denied the 1200 people of SCAT the chance to tell me what they did yesterday.

I think the key is this: I could quiz three group heads, and each head could say that he's one day down (one day behind schedule). One or two of the three might mean they were going to be down only one day three months from now. But one day down for the third man might mean he was going to miss by a month, or cause another group to miss by a month.

We constantly assessed the impact of the current position. LM is not production, it's development in the truest sense of the word; for each LM mission is a stepping-stone to the next one.

Let's consider two philosophies to explain the basic difference between development, and production. I can use the philosophy that says I want parts on the shelf: if something breaks, I can take it off the shelf and put it into play. This indicates a production program. But in development, the first time you have an anomaly you want to find out right away what failed, why it failed, and how you can fix it so it doesn't fail again. When you determine the cause of the failure, you then modify all the subsequent boxes ... And you have to be careful not to overhandle the box because you're likely to destroy its integrity.

How did you work out the timing of the deliveries for subsystems so that related pieces would be ready at the proper time?

A master schedule was the real heart of the timing. I like to pace parts coming in slightly ahead of my needs. One ahead of need—never two, just one. Installing one part triggers the scheduling system to produce the next one, or deliver the next one as the case might be. Of course, we had confidence in some of the hardware, and we'd go ahead and install it. But we couldn't do that with some of the active pieces like the radar, which we'd been working with constantly.

Now where this "one ahead of need" system gets to be significant is if the LM on the pad, with it maximum priority, lays a requirement on us for a part. We'd take it from the last vehicle on the line or, if it wasn't there, from the next vehicle up the line, and so on. We had three vehicles on test here at all times, and there were three vehicles at Cape Kennedy.

Our job, in addition to moving vehicles out, was to support the Cape. I could pull a part from the No. 1 vehicle in Bethpage, figuring that I had more latitude to employ a "work-around" than they had down there. We had 24 hours; at the Cape they had only one hour.

What would you avoid doing again?

Nothing really stands out in my mind. We're constantly improving the system; better management methods were spawned by an evolutionary process. Once we put the LM together we had to hook the ground support equipment to it and check it to see if it was functioning. We're constantly refining this check-out system. When I get smarter I expect to be able to check some major system, knowing full well that if I get certain parameters out of it all systems must be working.

The development program is, by nature, one that teaches us what to avoid doing the next time. For example, we took 80 days of electrical subsystem testing and cut it to 20 through better test logic and experience. If we were going to work on LM for 10 years we could check it out in a short time. Weight is important. If you knew some circuitry in a given component was
required by another one, you'd use that same circuit rather than repeat it.

**What is the cost differential between manned and unmanned space vehicles?**

The human being can handle emergenices. Neil Armstrong could probably have landed on a rock and got away with it. But he saw it so he came down and moved over away from it. Some of the unmanned Surveyors landed, and some didn't. The cost would depend mostly on what the vehicle is sent for. The cheapest way to go is in a manned LM with no backup system.

**What element of chance was there in designing the LM to land and take off from the moon?**

We weren't certain how large the landing gear should be. We asked ourselves: Does radar reflect and bounce off the lunar surface, or does it penetrate x amount of feet and then bounce back? Would there be dust on the surface or not? What was the composition of the surface, and how big a footprint did we need to support it? We were extremely conservative. We made the pads three feet plus. We learned later, of course, that the pads penetrated the lunar surface by 1/4 inch.

**What was your most serious problem in assembling and testing LM?**

Water glycol. How was it used? You have to get rid of heat. On earth you can turn on the fan. But in the vacuum of space you don't have this capability. The thermal characteristics in space are extreme. In the absence of any atmosphere, the sun side of the vehicle gets very hot and the shade side gets very cold.

If I want to cool a piece of hot radio equipment I can suck the heat off through a cold plate. Now I've got the heat in the cold plate, but I still have to get rid of it some way. So I transfer the heat from the cold plate into a hollow tube containing water glycol. Then it passes on through sublimators. I'll boil off some water and pass the heat off into the hard vacuum of space.

The main problem with glycol is that it forms a residue. Many times we'd be working up on top of the LM and break a line. If we spilled the glycol we had to keep washing the LM down with distilled water and wipe the surface with litmus paper to make absolutely sure that the residue was off. If we didn't get it off and the water hit the thermal-shielding blanket it would injure it. Then we'd have to take the blanket off and change it. A water glycol spill was serious because it took time to correct the mistake, and time was our most important commodity.

*(continued on next page)*
How were employees motivated to maintain the high reliability of the LM?

A staff can build an airplane and watch it fly, and then they can work on it some more and watch it fly again. But after a LM is finished, your staff will never see it again except on TV. This kind of engineering makes it difficult for management to motivate a man to admit he's made a mistake. We overcame this problem through a fortunate selection of people over a long period of time. We chose people who had a human desire to do a good job.

If one of them goofed, I would talk to him and ask him if he had learned something from the mistake. I would ask him what he was going to do the next time. I like to know why people do things. Many will give you "reasons" when they're really only trying to convey a thought. The key to good management is in trying to determine your staff's mental process rather than the words they're using. If you understand the mental process, you can begin to solve their problem.

In development, all I can really guarantee is that I will never make the same mistake twice. That guarantee automatically means that I will keep getting better.

Did NASA breathe down your neck?

Sure. Sometimes we expected it—sometimes we didn't. I'll say this: The relationship I enjoyed in SCAT was closer and on a higher technical plane than any I've ever known. The problem we had to solve forced a much closer relationship; we didn't just talk contract ties, we talked technical problems and solutions.

What kind of engineering talent are you interested in for the future?

That depends on the direction the company wants to take in the future. What is the post-Apollo space program?

Whatever it is, it will include an orbiting space station. Most of the cost right now is in getting from earth into orbit. Of course, we'd still have to get up to the station, but I could get there with a bulk of fuel and run back and forth from the moon many, many times before I'd have to go back to earth. I could always send the spacecraft back to an orbiting space station, repair it, and stick it back in orbit.

Three things will drive the space program: Performance, cost and time. Apollo definitely had time as its driver, but it also had to perform. Cost? No one knew how to land on the moon at a lower cost and still do it in this decade. Time? Post-Apollo programs will be willing to let time slide.

Do you have any regrets about the LM program?

One regret I have is that the descent stage is on the moon forever more, and the ascent stage is orbiting the moon. We'll never get them into the Smithsonian Institution for future generations to see. **
TI introduces an MSI parity generator/checker which has no equal.

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Generate variable-phase square waves this cheap, easy way

Converting a sine wave into a constant-amplitude square wave with adjustable phase shift is often necessary in such circuits as phase-sensitive detectors. A very satisfactory solution to this need, using inexpensive monolithic operational amplifiers and operating over the full 0° to 180° range, is shown in the figure.

Amplifier A1 is an inverting sine wave generator, and A2 is a cosine generator. The outputs are summed together by A3 with relative weighting controlled by P1. The input to A3 is thus a sinusoid whose amplitude and phase are determined by the amplitudes of the input sine and cosine waves. Because A3 is operated open-loop, it acts as a zero-crossing detector generating a constant amplitude square wave whose phase can be adjusted from -90° to +90°. The circuit can be used at any frequency from 10 Hz to 1 MHz by setting the reactance of C to 10 kΩ at the desired frequency.


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Transformer-less Meacham bridge employs IC op amp

An untuned version of the well-known Meacham Bridge crystal oscillator can easily be built with an IC op amp (Fig. 1). Unlike many other IC crystal "clocks," this one maintains the crystal drive at a constant level, thus improving the circuit's frequency stability.

A #344 lamp on the negative feedback side of the bridge acts as a nonlinear element (Fig. 2) to cause an increase in the negative feedback factor when the output increases. This operating point stabilization is similar to that used in most RC audio generators.

Since the lamp operates at about 100 Ω and the crystal series resistance at series-resonance is about 1000 Ω, it is not possible to make all of the bridge arms equal. This would improve the stability of the design.

By using a vacuum-mounted crystal, with a lower series resistance, the all-equal-arm aim could be approached. Also, as op amps with higher frequency cutoffs become available, AT-cut crystals above 1 MHz could be used in similar designs. These crystals have about 100 Ω of series resistance near 2.5 MHz and would seem to be naturals for such a Meacham Bridge.

Henry D. Olson, Research Engineer, Stanford Research Institute, Menlo Park, Calif.

Divide by 3, 5 or 10 with a minimum of hardware

Many digital applications require frequency dividers of factors other than the binary quantities 2, 4, 8, etc. There are several techniques for designing such dividers, and most of them require additional gates or components.

A technique that is more economical in the use of components utilizes the set and clear in-

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A minimum number of components is required for these 3:1 (A), 5:1 (B) and 10:1 (C) frequency dividers.

puts of J-K flip-flops. Minimum-hardware frequency dividers of this type for factors of 3, 5, and 10 are shown in the illustrations. These arrangements are suitable for either DTL or TTL J-K flip-flops.

Basil Ioannou and Carl Brunnett, Design Engineers, Picker Instruments, Cleveland, Ohio.

Send us those programing short cuts

In line with the impact that time-sharing terminals are having on the work of the electronic designer, ELECTRONIC DESIGN will now accept computer programs and subroutines as Ideas for Design. The only requirements are that the program perform some useful design calculation or function, that it be in a recognized computer language, and that it not exceed 60 lines.

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Lumped model improves CAD analysis of transmission lines

Most programs for the computer-aided design (CAD) of electrical circuits are limited to lumped-parameter circuit representations. However, it is often necessary to handle transmission lines with such programs. A lumped-parameter approximation (Fig. 1) is usually used for this purpose.

If the CAD program can handle controlled voltage and current sources, the six-element equivalent circuit of Fig. 2 can be used to advantage. It is an exact equivalent of a lossless
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Write: Du Pont Company, Room H 7297, Wilmington, Del. 19898.
transmission line of characteristic impedance $Z_0$ and delay $\tau$. The circuit is especially well-suited to steady-state analysis.

The four controlled sources are:

- $V_{g1} = V_2 e^{j\omega t}$
- $I_{g1} = I_2 e^{j\omega t}$
- $V_{g2} = V_1 e^{j\omega t}$
- $I_{g2} = I_1 e^{j\omega t}$

If it is desired to use the model for a transient analysis, the equations should be replaced by their time-domain forms:

- $V_{g1}(t) = V_2 (t - \tau)$
- $I_{g1}(t) = I_2 (t - \tau)$
- $V_{g2}(t) = V_1 (t - \tau)$
- $I_{g2}(t) = I_1 (t - \tau)$

Of course, for the transient analysis, the computer must remember the histories of $I_1, I_2, V_1$, and $V_2$ from $t - \tau$ up to the present time, $t$.


VOTE FOR 314

LEDs have advantages as constant brightness sources

A light emitting diode (LED) can replace the incandescent lamp and provide constant brightness references for industrial equipment. The circuit shown combines good stability and control with low cost. Furthermore, there is no need to resort to optical feedback. Control is attained to $\pm 3\%$ of the diode's brightness at ambient temperatures over the range from $+10^\circ$ to $+50^\circ$. Also, there is no need to compensate for aging or blackening as is the case with incandescent lamps.

The circuit relies on the stability inherent in solid state light sources as well as their long life. Unlike incandescent lights, light-emitting diodes do have thermal coefficients but they are predictable.

In the circuit shown, the thermistor, RT1, provides the necessary temperature compensation. The constant brightness circuit is driven from a regulated dc supply providing $+12$ volts. The forward current of diode D1 at ambient temperature is set by resistor $R_3$. Five percent tolerance resistors are sufficient for the 3 percent control this circuit provides.

The temperature characteristic of the light-emitting diode is improved at least one order of magnitude by the thermistor. The LED temperature coefficient is negative, and the diode's light output decreases with increases in temperature. However, increases in temperature also decrease the thermistor's resistance thus increasing the base drive of Q1 and, consequently, the diode current. As a result, the diode's light output remains constant over the temperature range. For falling temperatures, the forward current is lowered by the increase in the thermistor's resistance. In this manner, the current through the diode is reduced as the temperature decreases, and its light output again remains stable with temperature changes.

William Otsuka, Applications Manager, Monsanto Co., Cupertino, Calif.

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- **LOAD REGULATION:** Less than ± 0.01% plus 1mV output voltage change for a load current change equal to the current rating of the supply.
- **LINE REGULATION:** Less than ± 0.02% output voltage change for a change in voltage from 105 to 125V or 125 to 105V volts at any output voltage and current within rating.
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Solid state all the way. Two new instruments, an integrating dc voltmeter and a programmable 150-MHz universal counter, readout with numeric LED arrays, p. 94.


Power plus. New SCR has 450-A rating and PIV of 1200 V, p. 97.

Also in this section:

Digital $89 panel meter with 100% overranging is accurate to 1%, p. 94.
Solid-state lamp is bright red LED with 150-μW 10-mA output, p. 98.
Transistor-sized balanced mixer performs from 25 kHz to 500 MHz, p. 100.
Application Notes, p. 111 . . . New Literature, p. 112.
Digital voltmeter and counter readout with LED numerics


Founding a unique generation of electronic instruments, a new integrating digital dc voltmeter and a programmable 150-MHz universal counter incorporate solid-state light-emitting numerics for readout display. Establishing a true industry first, both new instruments employ seven-segment LED arrays to form the numeric symbols.

Dispelling the myth that solid-state displays will double instrument costs, model 200A integrating digital voltmeter offers a full five-digit readout for only $895. Other features include 20% overranging, automatic polarity selection, and remotely programmable pushbutton range selection.

The 200A measures dc voltages to 1000 V in four ranges, with resolutions of 100 µV to 100 mV. Its accuracy is ±0.01% of full scale, ±0.05% of reading, ±1 digit, whichever is smaller.

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Three-digit $250 meter trims size with LSI

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<td>No</td>
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<tr>
<td>Available through Electronic Distributors</td>
<td>Yes</td>
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</table>

COMPARE! OAK LOWEST in Cost Cost is 60% HIGHER Cost is 140% HIGHER

GET OAK QUALITY PLUS THE LOWEST PRICE IN THE INDUSTRY

For full details on the sub-miniature switch that does more, write today for Bulletin SP-299.

Oak Versatility

Ultra Compact Rotary Switch
1/2" does a man-sized job

Rugged rotary outperforms its nearest competitors by the widest of margins, gives you man-sized performance at a bargain price. Don't believe it? Compare:

<table>
<thead>
<tr>
<th>FEATURES</th>
<th>OAK 1/2&quot; SWITCH</th>
<th>BRAND &quot;A&quot; SWITCH</th>
<th>BRAND &quot;B&quot; SWITCH</th>
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<tr>
<td>Double-wiping contacts</td>
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Oak Versatility

Electronic Design 20, September 27, 1969
Probe lights up to check logic levels in a flash

Make contact with the new Kurz-Kasch Logic Probe... With the speed of light, you can visually trace pulses or test the logic levels of DTL, TTL and related circuits. Probe flashes “true” and “zero” logic readings by illuminating signal lamps in the end of the instrument. Like having a lab of test equipment at your fingertips. Indicates “infinity” too, identifying improper logic or a disconnection. Displays symmetrical wave forms by illuminating both lamps.

You’ll light up at the low user price of $29.97. The probe is used for testing, inspection, troubleshooting and circuit design. Fits in a shirt pocket; leads attach to unit being tested for power. Responds to systems from 3.75 to 6.5 vdc. Input impedance: 150 kΩ (logic “true”). Logic Probe is available through your local electronic distributor for immediate delivery, or for demonstration upon request. For additional information write Kurz-Kasch, Inc., Logic Instrument Division, 1421 S. Broadway, Dayton, Ohio 45401. (513) 223-8161.

INSTRUMENTATION

Digital panel meter is 0.05% accurate

Computer Products, Inc., 2709 N. Dixie Highway, P.O. Box 23849, Fort Lauderdale, Fla. Phone: (305) 565-9565. P&A: $328; 15 to 30 days.

Requiring a panel space of only 2.5 by 3 by 4.2 in., model DP400 digital panel meter can measure dc voltages with an accuracy of 0.05%, ±1/2 count, and a 4000-count resolution. It offers eight voltage ranges from 200 mV to 400 V and features an internally selectable decimal point position. Its temperature coefficient is 0.005%/°F.

CIRCLE NO. 253

High-power pulser doubles as supply

Instrument Research Co., P.O. Box 231, Lincoln, Mass. Phone: (617) 897-7647. P&A: $1175; stock to 4 wks.

Delivering pulse widths of 0.5 µs to 1 ms, model PM-2 high-power pulse generator can also be used as a dual-output power supply for driving 1-A loads at 0 to 2 kV. Output pulse risetimes range from 160 V/µs to 20 kV/µs at a repetition rate of 1 to 20,000 pulses per second. Power output is 45 W average or 2.25 kW peak at a duty cycle of 1 to 2%.

CIRCLE NO. 255

Dual-trace oscilloscope ends flicker and drift


With a new triggering device that eliminates trace flicker, the 1050 solid-state dual-channel oscilloscope maintains signal synchronization regardless of the vertical positioning of either trace. In addition, the new instrument ends the need to readjust trigger level to maintain the desired reference level after trace repositioning. Bandwidth is dc to 50 MHz.

CIRCLE NO. 254

Memory voltmeter includes recorder

Micro Instruments Co., 12901 Crenshaw Blvd., Hawthorne, Calif. Phone: (213) 679-8237.

Able to measure transients as small as 50-ns wide, a new memory voltmeter uses a built-in strip-chart recorder to permanently record all readings. Model 5201CR has a dual-shielded cabinet isolated to 1000 V and can measure up to 30 kV with an optional probe. The recorder uses 63-ft dry-process pressure-sensitive paper at a speed of 30 inches per hour. Speeds from 10 to 150 inches per hour are optional.

CIRCLE NO. 256

96

ELECTRONIC DESIGN 20, September 27, 1969
ICs & SEMICONDUCTORS

**SCR with 1200-V PIV carries up to 450 A**

Siemens America Inc., Empire State Building, 350 Fifth Ave., New York City. Phone: (212) 564-7674. P&A: $165; 6 to 8 wks.

Offering a maximum peak inverse voltage rating of 1200 V, a new silicon controlled rectifier can handle continuous currents as high as 450 A. Type SSi N 20 has a normal PIV of 1000 V that can be increased to 1200 V with the addition of factory-installed pass transistors. Maximum periodic peak reverse voltage is 2000 or 2400 V, depending on the PIV rating.

**CIRCLE NO. 257**

**LSI digital analyzer is on-chip computer**

Texas Instruments Inc., Components Group, P.O. Box 5012, Dallas, Tex. Phone: (214) 238-2011. P&A: $750; 8 wks.

Using large-scale integration techniques to fit 253 equivalent gates on a single slice of silicon, a new digital differential analyzer is a special-purpose incremental computer for the solution of differential equations. Model DRA 1001 employs high-speed TTL parallel logic and can operate with an input clock rate of 2 MHz. Power dissipation is approximately 2.1 W.

**CIRCLE NO. 258**

---

**Heat problems?**

**Give 'em the air...**

Condor. A new high performance high reliability propeller fan providing up to 575 cfm for a wide range of cooling applications. The compact design (10-inch diameter, 3.5-inch depth) and light weight make it easy to install in a variety of equipments. 6 models with different connectors add to its versatility.

The Super boxer's exclusive new aerodynamic impeller design provides exceptional output characteristics at high back pressures. Super dependability. 2 patented bearing designs are rated at 10 years life under normal conditions. Super versatility. Compact (4.687-inch square, 1.5-inch depth), mountable inside or outside an enclosure, to intake or exhaust. Accepts all standard Boxer accessories.

DC Boxer. The small module mounted integrally on the Boxer frame accepts DC and converts it to drive the Boxer’s AC motor. Does away with usual DC motor problems, such as brush wear, arcing contacts, short life, and RF noise. 8 models span the range of 12 to 38 VDC input. Cools heat sensitive equipment such as, TV cameras, sound systems, telephone equipment, etc. Accepts all standard Boxer accessories.

**with 3 new air-givers from IMC.**

Distributor stocked nationwide for immediate delivery. As are standard Boxers, MiniBoxers, Tandem Boxers, IMCair fans and IMCair centrifugal blowers.

A new 16-page catalog provides drawings, performance parameters and complete specifications for all our airmovers. It's available from IMC Magnetics Corp., New Hampshire Division, Route 16B, Rochester, N. H. 03867, Tel. (603) 332-5300.
High Voltage Silicon Rectifiers
Available in production quantities now!

<table>
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<tr>
<th>HIGH VOLTAGE RECTIFIERS*</th>
<th>1000V</th>
<th>1500V</th>
<th>2000V</th>
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</tbody>
</table>

*Available with fast recovery characteristic.

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**ICs & SEMICONDUCTORS**

**Solid-state lamp beams bright red**

General Electric Co., Miniature Lamp Dept., P.O. Box 2422, Cleveland, Ohio. Phone: (216) 266-2248. Price: $9.52.

Easily visible in room light, a new bright red solid-state lamp peaks at 700 nm and produces 150 µW at 10 mA. This gallium-phosphide light-emitting diode, type SSL-22, is a plastic-encapsulated device, one-quarter of an inch in diameter. It has an end-on candle-power of 1.5 millicandela.

**MOS memory kits control displays**

National Semiconductor Corp., 2975 San Ysidro Way, Santa Clara, Calif. Phone: (408) 245-4320. Price: $120 or $150.

Designed to generate control signals for digital data displays, two new MOS read-only memory kits (models SK0001 and SK0002) contain three 1024-bit memories that are preprogrammed to convert standard communications codes (ASCII) into raster scan or vertical scan control signals. Each kit stores 64 standard 5 by 7 alphanumeric symbols.

**Power transistors turn-on in 10 ns**

Solitron Devices, Inc., 1177 Blue Heron Blvd., Riviera Beach, Fla. Phone: (305) 848-4311. Availability: stock.

Packaged in the popular TO-60 case, a new line of fast-switching 10-A power transistors feature turn-on times of less than 10 ns. Series S0T6110-S0T6113 devices have a total switching time of less than 80 ns and a risetime of less than 10 ns. They offer planar construction and a typical unity-gain cross-over frequency of 500 MHz.

**Economy MSI circuit decodes and drives**

Fairchild Semiconductor, 313 Fairchild Drive, Mountain View, Calif. Phone: (415) 962-3563. P&A: $10 to $22; stock.

Able to drive Nixie tubes directly, a new low-cost one-of-ten decoder/driver accepts 8-4-2-1 binary coded decimal (BCD) inputs and provides ten mutually exclusive outputs. Called the MSI 9315, this medium-scale integrated circuit has stable high-voltage output characteristics that also make it ideal for driving miniature lamps, relays and other high-voltage devices.
Power Darlington gain up to 20,000

Motorola Semiconductor Products, Inc., P.O. Box 20924, Phoenix, Ariz. Phone: (602) 273-6900. Price: $7.95 or $11.95.

Two new npn silicon power Darlington amplifiers, types MJ3801 and MJ3802, feature a minimum dc current gain of 20,000 and 2000, respectively. Both monolithic devices have a maximum collector-emitter cutoff current of 10 µA and a minimum collector-emitter breakdown voltage of 80 V. Power dissipation is 40 W and continuous collector current is 10 A.

CIRCLE NO. 263

MOSFET multiplexer has leakage of 200 pA

Union Carbide Corp., Semiconductor Dept., P.O. Box 23017, 8888 Balboa Ave., San Diego, Calif. Phone: (714) 279-4500. Availability: stock.

Consisting of six MOS p-channel enhancement transistors on a single chip, the MULT6 (UC6410) multiplexer holds leakage current to less than 200 pA and capacitance to under 1 pF. Besides a voltage breakdown rating of -40 V, the device offers added gate protection so that it can be handled like any bipolar device with no detrimental static charge accumulation.

CIRCLE NO. 264

IC op amp trio fit on one chip

RCA/Electronic Components, 415 S. Fifth St., Harrison, N.J. Phone: (201) 485-3900. Price: $5.95.

Designated the CA3060A integrated operational transconductance amplifier, a new IC op amp array consists of three transconductance amplifiers on a single silicon die. The electrical characteristics of each amplifier are a function of the total current of that particular amplifier. The current flow in each amplifier is externally adjustable from 100 nA to 1 mA.

CIRCLE NO. 265

NOW! from RMC

"ACROSS-THE-LINE" U.L. LISTED DISCAPS

CROSS-THE-LINE LISTED DISCAPS

EIA Class

<table>
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<tr>
<th>NPO</th>
<th>N750</th>
<th>N1500</th>
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<td>3.9-20</td>
<td>15-35</td>
<td>15-67</td>
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<tr>
<td>470</td>
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<td>21-31</td>
<td>34-61</td>
<td>68-119</td>
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<tr>
<td>120-180</td>
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</table>

21-31
32-47
62-82
120-180

RMC now offers a complete line of ceramic disc capacitors fully approved by Underwriters Laboratories for the NEW "Across-The-Line" capacitor requirements. This approval is required of all capacitors utilized directly or indirectly across the power supply line.

This application is significantly different from the "Antenna Coupling and Line By pass" capacitor requirements of Underwriters Laboratories Subject 492, and the original RMC -U- capacitor type continues to be approved for those applications.

SPECIFICATIONS

**CAPACITANCE:** Within specified tolerance:
- Class I @ 1MC and 25°C
- Class II @ 1KC and 25°C

**CAPACITANCE TOLERANCES AVAILABLE:**
- Class I ± 5%, ± 10% or ± 20%
- Class II ± 20%, ± 80–90%

**WORKING VOLTAGE:** 150 VRMS @ 60 cycles (210 volts peak AC plus DC)

**POWER FACTOR:**
- Class I 1.5% max. at 1 MC
- Class II 2% max. less than 30 pf
- Class III 1.2% max. at 1KC

**INSULATION RESISTANCE:** Greater than 7500 Megohms @ 500 VDC

**TEMPERATURE COEFFICIENT:**
- Class I NPO N750 N1500
- Class II Z5U, Z5F

**FLASH TEST:** Per U.L. Sub. 492

**LIFE TEST:**
- Per U.L. Sub. 492

**INSULATION RESISTANCE AFTER HUMIDITY:**
- Greater than 1000 Megohms @ 300 VDC

**BODY INSULATION:**
- Durex phenolic–vacuum wax impregnated. Standard coating on leads 1/4 max. measured from tangent

**LEAD STYLES AVAILABLE:**
- Long lead–#20 AWG tinned copper

**DISCAP CERAMIC CAPACITORS:**

RADIO MATERIALS COMPANY
A DIVISION OF P. R. MALLORY & CO., INC.
GENERAL OFFICE: 4242 W. Bryn Mawr Ave., Chicago, Ill. 60646
Two RMC Plants Devoted Exclusively to Ceramic Capacitors
FACTORIES AT CHICAGO, ILL. AND ATTICA, INB.

INFORMATION RETRIEVAL NUMBER 60
Transistor-sized mixer covers audio-to-uhf band

Mini-Circuits Laboratory Div., Scientific Components Corp., 2913 Quentin Rd., Brooklyn, N.Y. Phone: (212) 252-8252. P&A: $75; stock to 3 wks.

Spanning the frequency range from audio to uhf with a bandwidth of greater than 14 octaves, a new microminiature double-balanced mixer operates from 25 kHz to 500 MHz with a typical noise figure of less than 6 dB. This tiny device, model ASK-2, features a volume of only 0.05 cubic inches, occupying approximately the same board space as a TO-5 transistor package.

In addition, the new mixer can perform a variety of functions. It can be used as a very fast switch, a current controlled attenuator, a balanced modulator, a frequency doubler or a phase detector.

Its precision internal components include miniature printed circuit boards, hot-carrier diodes and specially designed transmission-line transformers. The unit is also shielded against emi.

Intended for 50-Ω systems, the ASK-2 has a maximum conversion loss of 7 dB from 200 kHz to 200 MHz and 10 dB from 25 kHz to 500 MHz. Minimum isolation between its local oscillator and rf ports is 30 dB from 25 kHz to 30 MHz and 15 dB from 30 to 500 MHz. Isolation specifications are identical between the local oscillator and i-f ports.

Minimizing conversion loss and maximizing isolation, a new transistor-sized double-balanced mixer spans the frequency band of 25 kHz to 500 MHz. Occupying but 0.05 cubic inches, this microminiature device holds conversion loss to 10 dB maximum and keeps isolation to 15 dB minimum.
High-Q air capacitor trims from both ends

Voltronics Corp., West St., Hanover, N.J. Phone: (201) 887-1517. P&A: $6.75; stock.

Separately tunable from either end, a new sealed dual air trimmer capacitor has a minimum Q of 4000 at 100 MHz. Capacitance values on both ends range from 0.8 pF to more than 10 pF. Model V3043 has high resolution and completely linear tuning. Its temperature coefficient is 50 ± 50 ppm/°C and operating temperature range is -55 to +150°C.

CIRCLE NO. 267

Segmented readout slims down depth


Capable of forming numerals 0 to 9, a new line of seven-segment modular readouts measures only 1.3 in. deep. Series 100 units use incandescent bulbs rated at 150,000 hours for long life. Another feature is a viewing angle of 170° at distances up to 75 feet. Optional lens color variations are available.

CIRCLE NO. 268
3 ways to prevent numerals from "spotting" at 55°C.

Fill with freon... keep in shade... or specify Datavue* Indicator Tubes.

All Datavue tubes are rigorously tested to meet commercial and military specifications, produced for 200,000 hours of reliable operation. U.S.-made, they can cost less than $3.95 each.

Datavue tubes feature: straight, stiff leads for fast insertion; fully formed, high-brightness characters; rated for strobing operation; wide range of alphanumeric, decimals, special characters. More than 40 different sockets, including right-angle types, are available.

Call your Raytheon distributor or sales office. Raytheon Company, Industrial Components Operation, Quincy, Mass. 02169.

*Trademark of Raytheon Company
DATA PROCESSING

CRT display system interfaces easily

Applied Digital Data Systems, Inc.,
89 Marcus Blvd., Hauppauge, N.Y.
Phone: (516) 273-7799. Price: from $995.

Called a static raster display, a new CRT readout offers a design engineer a numeric/symbolic device that can use any standard 525-line TV monitor. Interfacing to any system is as simple as that for indicator tubes. Model SRD-100 accepts direct BCD input signals from any digital data source and converts them into a composite video signal.

CIRCLE NO. 273

Read-only memory stores data manually

United Telecontrol Electronics,
3500 Sunset Ave., Asbury Park,
N.J. Phone: (201) 988-0400.

Adapting a general-purpose random-access memory to simulate read-only memory storage, a new system offers three major features: stored data can be changed (or entered) manually, address and data output binary information is continuously displayed, and access is provided for loading from tape, drum, card or core storage. Complete programs can contain up to 4096 eight-bit words.

CIRCLE NO. 275

Rack-mount memory meets MIL specs

United Telecontrol Electronics,
3500 Sunset Ave., Asbury Park,
N.J. Phone: (201) 988-0400.

Housed in a rack-mount chassis for use in standard 19-in. relay racks, model 2199M core memory system meets MIL-E-16400, Class IV, Type 1 equipment specifications. It is dimensionally compact at 5-1/4 by 19 by 14-1/2 in. overall. Its 1 by 8 core stack plugs into keyed edge connectors, as do its solid-state IC printed-circuit cards.

CIRCLE NO. 274

Electrostatic copier standardizes printouts


Specifically designed for the computer industry, the new electrostatic desk-top office copier quickly reduces computer printouts by 25% and makes them standard business-form size. Copystat CR-75 takes an 11 by 14-in. printout original and produces an 8-1/2 by 11-in. copy in order to save filing space and handling time.

CIRCLE NO. 276
Analog Multiplier

The new M510 multiplier is a moderately priced, epoxy encapsulated transconductance multiplier featuring very fast operation with excellent linearity and temperature stability. The 5 MHz full output frequency and 10 MHz response suit the M510 to modulation, correlation, display systems, and other applications requiring fast performance.

- 5 MHz Full Output
- DC to 10 MHz
- 1% Accuracy
- 300 V/µs Slew Rate
- No Adjustments
- True 4 Quadrant
- 2 x 3 x 0.62 Inches

SPECIFICATIONS (25°C, ±15 V)
- Output functions: XY/10, X²/10
- Linearity X: 0.5% max
- Linearity Y: 0.25% max
- Input voltage X, Y: ±10 V to common
- Input impedance X, Y: 10 kΩ, common
- Output (short protected): ±10 V, 10 mA
- Output impedance, dc: 1Ω max
- Frequency response:
  - Small signal (-3 dB): 10 MHz min
  - Full output (20 V p-p): 5 MHz min
- Gain temperature stability: ±0.02%/°C max
- Offset temperature stability: ±1 mV/°C max
- Power requirement: ±15 Vdc, 50 mA
- Price (1-9): $295

CIRCLE NO. 277

Philbrick/Nexus Research, a Teledyne Co., Allied Drive at Route 128, Dedham, Mass. Phone: (617) 329-1600. P&A: $55; stock.

Providing full output at 1 MHz and a small-signal bandwidth of 10 MHz, a new FET-input hybrid operational amplifier features a slew rate in excess of 50 V/µs and a rolloff of 6 dB/octave. Model 1405 is a differential unit with maximum bias currents of 50 pA and an offset voltage drift of 25 µV/°C. Open-loop gain is 300,000 typical.

CIRCLE NO. 278

Analog multiplier spans 3-MHz band

Optical Electronics Inc., P.O. Box 11160, Tucson, Ariz. Phone: (602) 624-8358. P&A: $65; stock.

Providing a total error of 1% when used with four external adjustments, a new four-quadrant analog multiplier boasts a large-signal bandwidth of 1 MHz and a small-signal bandwidth of 3 MHz. Model 5805 has a dynamic range of ±10 V on both inputs and an input impedance of 10 kΩ. Maximum untrimmed offset voltage is ±2 V for the inputs and ±3 V for the output.

CIRCLE NO. 279

Modular $35 multiplier delivers 1% accuracy

INTRONICS

Modular $35 multiplier delivers 1% accuracy

INTRONICS M420

INTRONICS INC., 57 CHAPEL ST., NEWTON, MASS. 02158

INFORMATION RETRIEVAL NUMBER 65

CIRCLE NO. 278

Modular $35 multiplier delivers 1% accuracy

INTRONICS INC., 57 CHAPEL ST., NEWTON, MASS. 02158

INFORMATION RETRIEVAL NUMBER 65

CIRCLE NO. 278

Low-cost multipliers eliminate trimming


Performing without the need for external trimming, two new low-cost multipliers work directly from standard ±15-V dc supplies. Tailored for general-purpose multiplier applications, models 4095/15 and 4096/15 exhibit less than a ±0.1% error with a 1% variation in either power supply. Accuracy is ±2 and ±1%, respectively, including all errors.

CIRCLE NO. 279

Modular $35 multiplier delivers 1% accuracy

INTRONICS INC., 57 CHAPEL ST., NEWTON, MASS. 02158

INFORMATION RETRIEVAL NUMBER 65

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CIRCLE NO. 279
SOLID TANTALUM CAPACITORS
FOR HYBRID ICS "MICROCAP".

Capacitance exceeding 10,000 pico-
farads obtained despite miniature size.
"MICROCAP" features excellent heat
resistance, solderability and mechanical
strength comparable to conventional
discrete components, for easy use in
hybrid integrated circuits.

Specifications:
- Operating Temperature Range: -55°C to +85°C
- Standard Voltage Rating: 6.3, 10, 16, 20, 25, 35 VDC
- Standard Capacitance Value: .001 to 22 MF (E6 series)
- Standard Capacitance Tolerance: ±20% (M)

MATSUO'S other capacitors include:
- Metallized Polyester Film Capacitor:
  Type FNX-H, mylar wrapped.
- Solid Tantalum Capacitors:
  Type TAX hermetically sealed in metallic
case, Type TSX encased in metallic
case and sealed with epoxy resin.
  Type TSL encased in metallic case and sealed
  with epoxy resin.
- Polyester Film Capacitors:
  Type MFL, epoxy dipped, non-inductive.
  Type MFK, epoxy dipped, non-
  inductive. Type MXT, encased in
  plastic tube, non-inductive.

For further information, please write to:
MATSUO ELECTRIC CO., LTD.
Head Office: 3-5, 3-chome, Sanmon-cho, Toyonaka-shi, Osaka, Japan
Cable: "MATSUO" OSAKA
Telex: 523-4164 OSA
Tokyo Office: 7-3-chome, Nishi-Osaka, Shinagawa-ku, Tokyo

Hand-sized supply powers 500 gates

Semiconductor Circuits, Inc., 163
Merrimac St., Woburn, Mass.
Phone: (617) 935-5200. P&A:
$49.95; stock.

Mounting directly on PC boards,
a new small power supply puts out
5 V at 500 mA to drive as many
as 500 logic gates. Model P1.5.500,
an encapsulated unit measuring 2.5
by 3.5 by 0.875 in., operates from
an ac source of 105 to 125 V, 50
to 400 Hz. Its line regulation is
0.05% maximum, and ripple and
noise are less than 1 mV rms.

CIRCLE NO. 282
ANY voltage from 2.0 to 16.0 at the industry's LOWEST PRICES!

<table>
<thead>
<tr>
<th>Quantity</th>
<th>Price each</th>
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<td>.86</td>
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<td>5000 up</td>
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THE HI-RELIABLE!

No fragile nail heads. Silicon junction aligned between two, parallel, offset tantalum heat sinks...great lead tension strength. All welded and brazed assembly. High pressure molded package. Gold plated nickel-clad copper leads. Write or phone for Form 68-4 for complete rating data and other tolerance prices.

SCHAUER MANUFACTURING CORP. 4511 Alpine Avenue Cincinnati, O. 45242 Ph. (513) 791-3030

INFORMATION RETRIEVAL NUMBER 68

MICROWAVES & LASERS

S-band transistors boast 5-W output

TRW Semiconductors Inc., 14520 Aviation Blvd., Lawndale, Calif. Phone: (213) 679-4561. P&A: $100 to $290; stock.

Three new microwave transistors operate at 2 GHz with a source voltage of only 28 V. Type 2N5766, which has a 1-W output, can withstand an infinite VSWR at any phase angle. Type 2N5767 supplies a 2.5-W output and can withstand a 3:1 VSWR at any phase angle. Type 2N5768 with a 5-W output can also withstand a 3:1 VSWR at any phase angle.

CIRCLE NO. 283

Solid-state oscillator gives 10 mW at 5 GHz

Engelmann Microwave Co., Skyline Drive, Montville, N.J. Phone: (201) 334-5700.

Mechanically tunable over ±50 MHz of the center frequency with ±1% electronic tunability, a new miniature solid-state microwave oscillator develops an output power in excess of 10 milliwatts in the 5-GHz region. The unit has a frequency stability of 0.1% over a temperature range of 0 to 60°C and 0.4% over the range of -30 to +70°C.

CIRCLE NO. 284

Nine-digit counter spans 20 Hz to 3 GHz


Able to service the new communication and telemetry bands, a microwave frequency counter provides complete coverage from 20 Hz to 3 GHz on a single input. Model 970, which sells for $2250, offers a full nine-digit readout. This new microwave instrument also features a resolution of 0.1 Hz and a sensitivity of —7 dBm.

CIRCLE NO. 285

Activity monitor sees out to 12 GHz


Intended for a quick-look analysis of spectrum activity, the WJ-1140 microwave monitor provides a CRT display of signal activity in the frequency band from 1 to 12 GHz. Readily installed for airborne, shipboard or van applications, the new compact solid-state system is ideal for situations involving restricted volume.

CIRCLE NO. 286
PACKAGING & MATERIALS

Modular rfi enclosures protect out to 18 GHz

Signal Interference Control, Inc., 161 Gazza Blvd., Farmingdale, N.Y.

Using aluminum modular construction with a patented gasket retention groove, a new system of rfi enclosures provides a shielding effectiveness of 100 dB over a frequency range of 15 kHz to 18 GHz. The gasket retention feature permits the removal of all six panels, allowing high component density and reduced production costs. The modular construction provides the engineer with a custom enclosure.

CIRCLE NO. 287

Wash-away adhesive holds delicate parts

Remco Products, Inc., P.O. Box 145, Briarcliff Manor, N.Y. Phone: (914) 762-0685. P&A: $25/stick; stock.

Crystalbond 509, a thermoplastic polymer that is soluble in acetone, can be used as a temporary bond, wash-away adhesive, for holding delicate parts. It can bond crystals and ceramic substrates for slicing, grinding, dicing or polishing operations. The material, which comes in stick form, adheres to porous or non-porous metals and glass or ceramic surfaces.

CIRCLE NO. 288

Flexible PC mask withstands 500°F

EPD Industries Inc., Laboratories Div., 2055 E. 223rd St., Long Beach, Calif. Phone: (214) 775-7141.

Providing full-time protection for PC boards through all manufacturing and assembly operations, a new low-viscosity flexible masking compound withstands temperatures up to 300°F for one to three hours or 500°F for a sufficient time to allow flow soldering. TC-533 can remain on the protected area of the board until ready for test, then with a mild pulling action, can be removed without leaving a residue.

CIRCLE NO. 289

Precious-metal solders come ready to use

Dynalog, Inc., 7 Great Meadow Lane, Hanover, N.J. Phone: (201) 887-9270. Price: $15 for 2 oz (312), $60 for 1/2 oz (412)

Filled with a precious metal, two new epoxy solders, type 312 silver and type 412 gold, are single-component materials, ready to use as supplied. The 412 gold is recommended for applications where silver migration might present a problem. Both products can be used for lead terminations, chip bonding, conductive paths, circuit assembly and repairs.

CIRCLE NO. 290

The greater the variety, the more flexibility you have when designing a miniature ALCOSWITCH into your quality equipment. The broad ALCOSWITCH line offers over a hundred different models.

Miniature Toggle Switches are available in five series including high performance waterproof types, locking handles, and switches with 15/32" bushings.

Miniature Push Button Switches include the new illuminated series, top-grade SPDT-DPDT 4PDT types, 15/32" bushings, and economy priced miniatures.

Whether you need one or one thousand switches you have a greater variety to choose from when you specify an ALCOSWITCH.

SEND FOR \*PAGE ALCOSWITCH CATALOG

ALCO ELECTRONIC PRODUCTS, INC.
Lawrence, Massachusetts 01843
New 25 Amp Power Relay has a 10 Amp Auxiliary Switch

Deltrol Controls’ New 900-1C Power Relay has a snap-action, 10 amp, 1/3 hp, SPDT auxiliary switch for switching signal lights or other sensing devices—and heavy duty, 25 amp, contacts for motor control, heater loads, or other power switching applications. You get two relays in one! All 900 Series Power Relays are rugged, built to last: coil terminal breakage is eliminated because coils terminate on the base, long-life, extra heavy gold-flashed, silver cadmium oxide contacts, 9.5VA nominal power, recognized under the UL & CSA component listing. Get the complete story by asking for Engineering Bulletin No. 1372.

Soldering systems use IR radiation

Argus Engineering Co., Hopewell, N. J. Phone: (609) 466-1677. P&A: from $2750; 60 days.

Series FS flat flexible cable soldering systems make all types of soldered connections rapidly and reliably by using focused infrared heating techniques. Entire rows or groups of joints are soldered simultaneously; the heater never touches the workpiece. A complete row of terminations, often as many as 200 joints, typically requires less than five seconds to complete.

Blasting system cleans small parts

Industrial Products Div., Inland Manufacturing Co. 1108 Jackson St., Omaha, Nebraska. Phone: (402) 342-1108.

Designed for precision blasting operations or cleaning of small intricate parts, the Mini-Blaster is a tabletop console that provides safe and clean blasting. It can be used on a production basis or for intermittent use throughout the day. The new unit consists of an all-steel blasting cabinet with an inside lighting fixture and a 7 by 18-in. plexi-glass window. The cabinet aids in eliminating environmental contamination.

Vacuum pickup tool has metal/Teflon tips

Labtron Scientific Corp., 100 Smith St., Farmingdale, N.Y. Phone: (516) 293-4898. Price: $20.

Able to pick up and hold small and fragile items, a new vacuum forceps-like tool uses interchangeable tips of either metal or Teflon in sizes from 0.5 to 6 mm. Called Lab-69, the vacuum pickup connects directly to any convenient vacuum source, a vacuum pump or aspirator, and utilizes fine screen wires to prevent the tiny particles from being sucked into the tool. It has a comfortable pen-like shape.
Evaluation Samples

Terminal insulator

AMP/T post-insulating pods are one-piece molded nylon insulators designed for snap-on installation on crimped FASTON terminals. The pods are hand-installed and can be quickly and easily removed for terminal inspection or replacement. Each pod has a molded-in locking member which permits their use on standard non-insulated terminals allowing one terminal type to be used for either insulated or non-insulated applications. Once installed, the pod completely surrounds the terminal and provides a chamfered lead-in at the entry surface which eliminates misalignment due to terminal-to-tab misalignment. AMP Inc.

Module markers

Self-sticking Poly-Plates offer a fast, neat, permanent method of identifying electronic modules used in computers, switch gear, circuitry and equipment. All necessary data, including trademarks, wording, codes and colors, is sub-surface printed on these polyester labels. Individual specifications may include a special write-in area that accepts variable information written with a ballpoint pen, typewriter or other instrument. W. H. Brady Co.
Ye olde memoirs

Interested in knowing how it really was way back when? Then “Memoirs of the Royal Society” is an absolute must for your insight into those days of yore. Learn how the great minds of the seventeenth and eighteenth century mapped out the design details of a machine for changing the air of a room of sick people. In-depth descriptions and precise equations reinforce the text. IMC Magnetics Corp., Eastern Div.

CIRCLE NO. 342

Resins folder

Epoxy, polyurethane and polystyrene resins as well as other stycaot materials are charted in a new six-page folder. The properties tabulated include viscosity, cure temperature, yield strength, thermal conductivity, thermal expansion, service temperature, dielectric strength and volume resistivity. Emerson & Cuming, Inc.

CIRCLE NO. 343

PC coating selector

A new selector guide for printed circuit coatings lists the major characteristics of eight different printed circuit coatings, five urethane and three epoxy. Four of the coatings are available in military versions to meet the requirements of MIL-I-46058B. Differences between the various coatings, such as film thickness, pot life and shock resistance are treated in an easy-to-use reference form. The Dexter Corp., Hysol Div.

CIRCLE NO. 344

Design Aids

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Giving an account of the undertakings, studies, and
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Order of Time regularly observed, with a translation of the
INFORMATION RETRIEVAL NUMBER 73

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IN THE SPACE OF

Where your equipment or system requires more than one regulated DC output, consider Acopian duals. They consist of two independent regulated power supplies housed in a single module. You can select two like outputs (such as for op amps) or any of 80,000 combinations of different outputs.

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INFORMATION RETRIEVAL NUMBER 73

Annual Reports

Learn how to read annual reports in “How to investigate a company.” For a copy, circle no. 474.

Associated Spring Corp., 18 Main St., Bristol, Conn.: springs, connectors, data processing; net sales, $108,562,691; net income, $5,366,371; assets, $40,264,106; liabilities, $10,561,734.

CIRCLE NO. 345

Diebold, Inc., Canton, Ohio: closed circuit television, vaults, surveillance systems; net sales, $100,486,355; net income, $4,746,774; assets, $52,259,906; liabilities, $12,026,620.

CIRCLE NO. 346

Emerson Electric Co., 8100 Florida St., St. Louis, Mo.: control, electronic and environmental systems and components; net sales, $522,046,000; net earnings, $39,458,000; assets, $237,078,000; liabilities, $54,314,000.

CIRCLE NO. 347

Farinon Electric, 935 Washington St., San Carlos, Calif: microwave and communication systems and components; net sales, $8,539,931; net income, $751,381; assets, $4,349,537; liabilities, $2,645,850.

CIRCLE NO. 348

Loral Corp., 688 White Plains Rd., Scarsdale, N.Y.: defense, electronic and aerospace systems and components; net sales, $50,005,285; net income (loss) $1,444,025; assets, $23,097,860; liabilities, $8,596,758.

CIRCLE NO. 349

Viking Industries, Inc., 21001 Nordhoff St., Chatsworth, Calif.: connectors, sockets and fittings; net sales, $6,874,477; net income, $378,395; assets, $2,442,275; liabilities, $836,333.

CIRCLE NO. 350
Standards guide

The Standards Institute is now offering its 1969 catalog of USA Standards and international recommendations. Some 600 USA Standards and 300 international recommendations have been added to the new 112-page edition. It lists 3600 USA Standards approved by the Institute and 1350 international recommendations of the International Organization for Standardization (ISO), International Electromechanical Commission (IEC), International Commission on Rules for the Approval of Electrical Equipment (CEE), and Pan American Standards Commission (COPANT). United States of America Standards Institute.

CIRCLE NO. 351

Cryogenics

Entitled “Cryogenic Data Reference,” a new comprehensive 20-page booklet defines the science of cryogenics and provides a brief review of the state of the art. This review covers the alteration of matter at low temperatures, biological and medical applications, superconductivity, ultraprecise instrumentation, space simulation and metallurgy. Included are tables listing the cryogenic properties of gases, equivalent temperatures, and property equivalent of gases, as well as a temperature conversion table for Fahrenheit, Kelvin, Centigrade, and Rankine. Union Carbide Corp., Linde Div.

CIRCLE NO. 352

Permanent magnets

Temperature effects on permanent magnets are analyzed in a 12-page issue of “Applied Magnetics,” Vol. 16, No. 1. The booklet tells how increasing temperature varies saturation magnetization and intrinsic coercive force for both alnico and ceramic magnets. The three temperature effects—reversible, irreversible and material—and their relationship to remanence are also explained and compared on temperature and second quadrant B-H curves. Indiana General Corp., Magnet Div.

CIRCLE NO. 353

Connector terms

Defining nearly 150 phrases, an eight-page glossary of connector terms eliminates the confusion that exists in connector terminology. Generic words such as barrel, ferrule, insert and ramp have specific meanings in terms of connector design. The glossary provides a solid basis for communication between engineers and the connector manufacturer. Cinch Distributor Div., Cinch Manufacturing Co.

CIRCLE NO. 354

Ku-band multipliers

A six-page application note gives design details for a practical times-eight single-stage step-recovery-diode frequency multiplier with typical maximum output power of 75 mW at 16 GHz. The literature includes references and a discussion of how to modify the design to meet other performance requirements. Hewlett-Packard Co.

CIRCLE NO. 355

Application Notes
New Literature

Heathkit catalog
Said to illustrate the world's largest selection of electronic kits, the new 116-page Heathkit catalog shows over 300 kits for every budget and interest. Included are stereo and high-fidelity components, ham-radio equipment, marine gear, test equipment, shortwave and citizen-band receivers, photographic aids and educational kits. Heath Co.

Digital multimeter
A new eight-page brochure describes a precision five-digit multimeter with a sixth-digit 10% overrange capability. The instrument has an accuracy of 0.0025%. It can measure such parameters as millivolts, volts, resistances and ratios. It features single-point calibration and a one-year calibration warranty. Dana Laboratories, Inc.

Microwave products
Comprising a large comprehensive presentation, a new 48-page catalog contains complete information on a line of microwave components and equipment. Described are ferrite circulators and isolators, dummy loads, antenna horns, coaxial attenuators and terminations and slotted lines. Lectronic Research Laboratories, Inc.

MIL-type capacitors
Said to condense 202 pages of generic military specifications and 69 pages of data sheets into 28 pages of ready reference information, a new catalog details both solid and wet MIL-type tantalum capacitors. Covered are MIL-C-39003, MIL-C-3966C, MIL-C-3966D and MIL-C-39006, along with part numbers, ratings, environmental requirements and failure rate information. NCI Capacitor Div.

Digital display terms
Containing a glossary of terms common to digital display devices, a new two-page technical bulletin defines analog and digital signals, sample rate, common-mode rejection, print command and other terms. The glossary is part of a data sheet on a digital panel meter that offers a 0.1% full-scale accuracy, a 3-1/2-digit display, and automatic polarity and overranging. The Triplett Corp.

Equipment rental
Listing a wide variety of electronic equipment available for short-term rental, a new 48-page catalog includes: oscilloscopes, counters, recorders, signal generators, digital voltmeters, power supplies, other types of general purpose test equipment, as well as small computers and computer peripheral equipment. Rentronix Div. of Inmark, Inc.

Electronics aerosols
Covering chemical products for electronic equipment servicing, a new eight-page catalog describes tuner sprays, contact and control cleaners, insulating sprays, lubricants and circuit coolers. Chemtronics Inc.

IC test sockets
Covering a complete line of sockets and carriers for testing integrated circuits, a new catalog also describes a variety of accessory and universal-type sockets. The products are divided into twelve different categories, including a line for MSI and LSI devices. Robinson-Nugent, Inc.

Transformer guide
Comprehensive information on how to select and apply a wide variety of dry-type transformers is presented in a new 28-page booklet. It contains application and ordering guidelines for general-purpose, buck-boost, distribution, and machine tool transformers. Integral distribution centers and ac voltage stabilizers are also featured. The brochure also provides application information on voltage-tap arrangements. General Electric Co.

Fluidic devices
Pneumatic logic and fluidic components are described in a new catalog. Included is detailed information on interface valves, five-port sockets, bistable and monostable amplifiers, proportional amplifiers, fluid diodes, and miniature pneumatic valves. Applications, specifications, operating characteristics, dimensions, ordering information and prices are fully discussed. Parker Hannifin Corp., Pneumatic Div.
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INFORMATION RETRIEVAL NUMBER 84

NEW LITERATURE

Motor catalog
An eight-page catalog on computerized precision instrument motors shows how high-speed digital computer curves relate motor rpm to shaft torque, power factor, percent efficient and other variables to achieve optimum predictable motor performance. Presented are commercial and Mil-spec motors with helpful data on selection and performance. There is also tabular engineering information on polarized synchronous, hysteresis synchronous and induction motors. McLean Engineering Laboratories.

CIRCLE NO. 367

Rugged connectors
FRF firewall connector line is described in a new 12-page catalog. This high-temperature environmental-resistant series meets the requirements of MIL-C-5015, Class K subsonic, supersonic and hypersonic aerospace applications. ITT Cannon Electric.

CIRCLE NO. 368

Resistor array
Specifically designed, with resistance values and terminal arrangements, to be used with Fairchild's uA722 10-bit current source, a cermet thick-film binary resistor array is the subject of a four-page brochure. Schematic drawings, performance characteristics and specifications are provided for the array. Beckman Instruments, Inc., Helipot Div.

CIRCLE NO. 369

Multiplexer system
A time division multiplexer with a wide variety of remote terminal units is described in a comprehensive brochure. Typical applications of the multiplexer include computer systems devoted to time sharing. The system, which can be easily expanded, can transmit up to 38 full-duplex channels of data over a single 3-kHz voice-grade circuit. Tel-Tech Corp.

CIRCLE NO. 370
Free Career Inquiry Service
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**Employment History** - present and previous employers

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**Additional Training** - non-degree, industry, military, etc.

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**Professional Societies**

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910 911 912 913 914 915 916 917 918 919 925

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Advertisers' Index

Acopian Corporation ...................................... 110
Airpax Electronics, Incorporated ................................ 65
Alco Electronic Products, Inc. ................................ 103, 105, 107
Allen-Bradley Co. ........................................ 41, 43
Applied Digital Data Systems, Inc. .......................... 8, 9
Atlas Fibre Co. ............................................. 119
Barnes Corporation ......................................... 117
B & B Electronics Corp. ..................................... 114
Bourns, Inc. ................................................. 53
Burroughs Corporation ....................................... 20
Clevite Corporation, Burgess Battery Division ............. 64
Clevite Corporation, Piezoelectric Division .................. 119
Components, Inc. ........................................... 10
Cortina Glass Works, Electronic Products Division ........ 34, 35
Cunningham Corporation, Subsidiary of Gleason Works ...... 33
Curtin Development & Mfg. Co. ................................ 116
Cutler-Hammer .............................................. 92
Dale Electronics, Inc. ....................................... Cover III
Data Laboratories, Inc. ....................................... 47
Deltol Controls ............................................. 108
Du Pont Company ........................................... 90
Elco Corporation ............................................. 11
Electronic Design .......................................... 116
Fairchild Controls, A Division of Fairchild Camera and Instrument Corporation ................................ 23
General Electric Company, Lamp Metals & Components Dept. ................................ 119
General Radio Company ..................................... 1
Guardian Electric Manufacturing Company .................. 15
Hayden Book Company, Inc. ................................ 116
Hewlett-Packard ............................................ Cover II, 24
Honeywell, Computer Control Division ....................... 49
Honeywell Test Instrument Division .......................... 56 A-B, 57
IMC Magnetics Corporation .................................. 97
Ideal Precision Meter Co. Inc. ................................ 56
Intel Corporation ............................................ 60, 61
Intromics, Incorporated ...................................... 104
Isomet Corporation .......................................... 73
JFD Electronics Corp./Components Division .................. 100
Kurz-Kasch, Inc. ............................................. 96
Lockheed Electronics Company ................................ 88
McCoy Electronics Company ................................ 86
Matsuo Electric Co., Ltd. .................................... 105
Monsanto Company .......................................... 109
Motorola Communications & Electronics Inc. .................... 95
Motorola Semiconductors Products, Inc. ...................... 4, 5, 16, 50
National Semiconductor Corporation .......................... 12, 13
New Product Engineering, Inc. ................................ 54, 55
Nynstra, Inc. ................................................ 31
Oak Manufacturing Co. ....................................... 95
Power/Mate Corp. ........................................... 92
RCA Electronic Components and Devices ...................... 29, Cover IV
Radio Materials Company .................................... 99
Raytheon Company, Industrial Components Operation ....... 102
Raytheon Company, Semiconductor Operation .................. 62, 63
Schauer Manufacturing Corp. ................................ 106
Signetics Corporation ........................................ 27, 84
Simpson Electric Company .................................... 101
Spectroline Electronics Corporation ......................... 66
Sprague Electric Company .................................... 18
Systron-Donner Corporation .................................. 2
TRW Semiconductors, Inc. .................................... 58
Tektronix, Inc., Oscilloscopes ................................ 16 A-B-C-D
Telenic Engineering, A Division of Telenic Industries, Inc. ...... 117
Tenneyl Engineering, Inc. ..................................... 95
Texas Instruments Incorporated, Components Group ......... 6, 7, 83
Tektronix Instrument Corporation ............................. 17
Vario Inc. ..................................................... 98
Welch Allyn, Inc. ............................................ 14
Y-Labs .............................................................. 118

Career Advertising

Sperry Flight Systems Division ................................ 114
Tektronix, Inc. ............................................... 113

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<table>
<thead>
<tr>
<th>Category</th>
<th>Page</th>
<th>IRN</th>
</tr>
</thead>
<tbody>
<tr>
<td><strong>Components</strong></td>
<td></td>
<td></td>
</tr>
<tr>
<td>amplifiers, Darlington</td>
<td>99</td>
<td>263</td>
</tr>
<tr>
<td>analyzer, digital</td>
<td>97</td>
<td>258</td>
</tr>
<tr>
<td>capacitor, trimmer</td>
<td>101</td>
<td>267</td>
</tr>
<tr>
<td>capacitors, military (NL)</td>
<td>112</td>
<td>359</td>
</tr>
<tr>
<td>capacitors, monolithic</td>
<td>102</td>
<td>269</td>
</tr>
<tr>
<td>connector terms (AN)</td>
<td>111</td>
<td>354</td>
</tr>
<tr>
<td>decoder/driver, Nixie</td>
<td>98</td>
<td>262</td>
</tr>
<tr>
<td>design memoirs (DA)</td>
<td>110</td>
<td>342</td>
</tr>
<tr>
<td>display terms (NL)</td>
<td>112</td>
<td>360</td>
</tr>
<tr>
<td>fluidic devices (NL)</td>
<td>112</td>
<td>365</td>
</tr>
<tr>
<td>Heathkit catalog (NL)</td>
<td>112</td>
<td>356</td>
</tr>
<tr>
<td>insulator, terminal (ES)</td>
<td>109</td>
<td>340</td>
</tr>
<tr>
<td>LED, red</td>
<td>98</td>
<td>259</td>
</tr>
<tr>
<td>magnets, permanent (AN)</td>
<td>111</td>
<td>353</td>
</tr>
<tr>
<td>markers, module (ES)</td>
<td>109</td>
<td>341</td>
</tr>
<tr>
<td>memories, display</td>
<td>98</td>
<td>260</td>
</tr>
<tr>
<td>memory, core</td>
<td>103</td>
<td>275</td>
</tr>
<tr>
<td>memory system</td>
<td>103</td>
<td>275</td>
</tr>
<tr>
<td>microwave products (NL)</td>
<td>112</td>
<td>358</td>
</tr>
<tr>
<td>mixer, balanced</td>
<td>100</td>
<td>266</td>
</tr>
<tr>
<td>motors (NL)</td>
<td>114</td>
<td>367</td>
</tr>
<tr>
<td>multiplexer</td>
<td>99</td>
<td>264</td>
</tr>
<tr>
<td>multiplexer system (NL)</td>
<td>114</td>
<td>370</td>
</tr>
<tr>
<td>multiplier, analog</td>
<td>104</td>
<td>278</td>
</tr>
<tr>
<td>multiplier, low-cost</td>
<td>104</td>
<td>280</td>
</tr>
<tr>
<td>multipliers, economy</td>
<td>104</td>
<td>279</td>
</tr>
<tr>
<td>multipliers, Ku-band (AN)</td>
<td>111</td>
<td>355</td>
</tr>
<tr>
<td>op amp array</td>
<td>99</td>
<td>265</td>
</tr>
<tr>
<td>photocell, CdS</td>
<td>102</td>
<td>271</td>
</tr>
<tr>
<td>power supply, logic</td>
<td>105</td>
<td>282</td>
</tr>
<tr>
<td>readout, segmented relays, reed</td>
<td>101</td>
<td>268</td>
</tr>
<tr>
<td>resistor array (NL)</td>
<td>114</td>
<td>368</td>
</tr>
<tr>
<td>resistors, chip</td>
<td>102</td>
<td>272</td>
</tr>
<tr>
<td>SCR, 1200-V 450-A</td>
<td>97</td>
<td>257</td>
</tr>
<tr>
<td><strong>Data Processing</strong></td>
<td></td>
<td></td>
</tr>
<tr>
<td>copier, electrostatic</td>
<td>103</td>
<td>276</td>
</tr>
<tr>
<td>display system, CRT</td>
<td>103</td>
<td>273</td>
</tr>
<tr>
<td>display terms (NL)</td>
<td>112</td>
<td>360</td>
</tr>
<tr>
<td>equipment rental (NL)</td>
<td>112</td>
<td>361</td>
</tr>
<tr>
<td>fluidic devices (NL)</td>
<td>112</td>
<td>365</td>
</tr>
<tr>
<td>memory, core</td>
<td>103</td>
<td>274</td>
</tr>
<tr>
<td>memory system</td>
<td>103</td>
<td>275</td>
</tr>
<tr>
<td>multiplexer system (NL)</td>
<td>114</td>
<td>370</td>
</tr>
<tr>
<td>standards guide (AN)</td>
<td>111</td>
<td>351</td>
</tr>
<tr>
<td><strong>ICs &amp; Semiconductors</strong></td>
<td></td>
<td></td>
</tr>
<tr>
<td>analyzer, digital</td>
<td>97</td>
<td>258</td>
</tr>
<tr>
<td>amplifiers, Darlington</td>
<td>99</td>
<td>263</td>
</tr>
<tr>
<td>decoder/driver, Nixie</td>
<td>98</td>
<td>262</td>
</tr>
<tr>
<td>LED, red</td>
<td>98</td>
<td>259</td>
</tr>
<tr>
<td>memories, display</td>
<td>98</td>
<td>260</td>
</tr>
<tr>
<td>multiplexer</td>
<td>99</td>
<td>264</td>
</tr>
<tr>
<td>op amp array</td>
<td>99</td>
<td>265</td>
</tr>
<tr>
<td>resistor array (NL)</td>
<td>114</td>
<td>368</td>
</tr>
<tr>
<td>resistors, chip</td>
<td>102</td>
<td>272</td>
</tr>
<tr>
<td>SCR, 1200-V 450-A</td>
<td>97</td>
<td>257</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Category</th>
<th>Page</th>
<th>IRN</th>
</tr>
</thead>
<tbody>
<tr>
<td>sockets, IC</td>
<td>112</td>
<td>362</td>
</tr>
<tr>
<td>standards guide (AN)</td>
<td>111</td>
<td>351</td>
</tr>
<tr>
<td>transistors, power</td>
<td>98</td>
<td>261</td>
</tr>
<tr>
<td>transistors, S-band</td>
<td>106</td>
<td>283</td>
</tr>
<tr>
<td><strong>Instrumentation</strong></td>
<td></td>
<td></td>
</tr>
<tr>
<td>counter, frequency</td>
<td>106</td>
<td>285</td>
</tr>
<tr>
<td>counter, LED-display</td>
<td>94</td>
<td>250</td>
</tr>
<tr>
<td>cyrogenics (AN)</td>
<td>111</td>
<td>352</td>
</tr>
<tr>
<td>display system, CRT</td>
<td>103</td>
<td>273</td>
</tr>
<tr>
<td>display terms (NL)</td>
<td>112</td>
<td>360</td>
</tr>
<tr>
<td>equipment rental (NL)</td>
<td>112</td>
<td>361</td>
</tr>
<tr>
<td>generator, pulse</td>
<td>96</td>
<td>255</td>
</tr>
<tr>
<td>Heathkit catalog (NL)</td>
<td>112</td>
<td>356</td>
</tr>
<tr>
<td>meter, digital panel</td>
<td>94</td>
<td>252</td>
</tr>
<tr>
<td>meter, digital panel</td>
<td>96</td>
<td>253</td>
</tr>
<tr>
<td>microwave products (NL)</td>
<td>112</td>
<td>358</td>
</tr>
<tr>
<td>monitor, signal</td>
<td>106</td>
<td>286</td>
</tr>
<tr>
<td>multimeter, digital</td>
<td>94</td>
<td>251</td>
</tr>
<tr>
<td>multimeter, digital (NL)</td>
<td>112</td>
<td>357</td>
</tr>
<tr>
<td>multiplexer system (NL)</td>
<td>114</td>
<td>370</td>
</tr>
<tr>
<td>oscilloscope</td>
<td>96</td>
<td>254</td>
</tr>
<tr>
<td>standards guide (AN)</td>
<td>111</td>
<td>351</td>
</tr>
<tr>
<td>standards guide (AN) transistors, S-band</td>
<td>106</td>
<td>283</td>
</tr>
<tr>
<td><strong>Modules &amp; Subassemblies</strong></td>
<td></td>
<td></td>
</tr>
<tr>
<td>amplifiers, Darlington</td>
<td>99</td>
<td>263</td>
</tr>
<tr>
<td>analyzer, digital</td>
<td>97</td>
<td>258</td>
</tr>
<tr>
<td>decoder/driver, Nixie</td>
<td>98</td>
<td>262</td>
</tr>
<tr>
<td>display system, CRT</td>
<td>103</td>
<td>273</td>
</tr>
<tr>
<td>display terms (NL)</td>
<td>112</td>
<td>360</td>
</tr>
<tr>
<td>fluidic devices (NL)</td>
<td>112</td>
<td>365</td>
</tr>
<tr>
<td>Heathkit catalog (NL)</td>
<td>112</td>
<td>356</td>
</tr>
<tr>
<td>markers, module (ES)</td>
<td>109</td>
<td>341</td>
</tr>
<tr>
<td>memories, display</td>
<td>98</td>
<td>260</td>
</tr>
<tr>
<td>memory, core</td>
<td>103</td>
<td>275</td>
</tr>
<tr>
<td>memory system</td>
<td>103</td>
<td>275</td>
</tr>
<tr>
<td>microwave products (NL)</td>
<td>112</td>
<td>358</td>
</tr>
<tr>
<td>mixer, balanced</td>
<td>100</td>
<td>266</td>
</tr>
<tr>
<td>multiplexer</td>
<td>99</td>
<td>264</td>
</tr>
<tr>
<td>multiplexer system (NL)</td>
<td>114</td>
<td>370</td>
</tr>
<tr>
<td>multiplier, analog</td>
<td>104</td>
<td>278</td>
</tr>
<tr>
<td>multiplier, low-cost</td>
<td>104</td>
<td>280</td>
</tr>
<tr>
<td>multipliers, economy</td>
<td>104</td>
<td>279</td>
</tr>
<tr>
<td>multipliers, Ku-band (AN)</td>
<td>111</td>
<td>355</td>
</tr>
<tr>
<td>op amp, FET</td>
<td>104</td>
<td>277</td>
</tr>
<tr>
<td>op amp, FET</td>
<td>105</td>
<td>281</td>
</tr>
<tr>
<td>op amp array</td>
<td>99</td>
<td>265</td>
</tr>
<tr>
<td>oscillator, 5-GHz</td>
<td>106</td>
<td>284</td>
</tr>
<tr>
<td>power supply, logic</td>
<td>105</td>
<td>282</td>
</tr>
<tr>
<td>readout, segmented relays, reed</td>
<td>101</td>
<td>268</td>
</tr>
<tr>
<td>relays, reed</td>
<td>102</td>
<td>270</td>
</tr>
<tr>
<td><strong>New Literature</strong></td>
<td></td>
<td></td>
</tr>
<tr>
<td>aerosols, electronics</td>
<td>112</td>
<td>362</td>
</tr>
<tr>
<td>capacitors, military</td>
<td>112</td>
<td>359</td>
</tr>
<tr>
<td>connectors</td>
<td>114</td>
<td>368</td>
</tr>
<tr>
<td>display terms (NL)</td>
<td>112</td>
<td>360</td>
</tr>
<tr>
<td>equipment rental</td>
<td>112</td>
<td>361</td>
</tr>
<tr>
<td>fluidic devices (NL)</td>
<td>112</td>
<td>365</td>
</tr>
<tr>
<td>Heathkit catalog (NL)</td>
<td>112</td>
<td>356</td>
</tr>
<tr>
<td>microwave products (NL)</td>
<td>112</td>
<td>358</td>
</tr>
<tr>
<td>modules, Ku-band (AN)</td>
<td>111</td>
<td>355</td>
</tr>
<tr>
<td><strong>Application Notes</strong></td>
<td></td>
<td></td>
</tr>
<tr>
<td>connector terms</td>
<td>111</td>
<td>354</td>
</tr>
<tr>
<td>cyrogenics</td>
<td>111</td>
<td>352</td>
</tr>
<tr>
<td>magnets, permanent</td>
<td>111</td>
<td>353</td>
</tr>
<tr>
<td>multipliers, Ku-band standards guide (AN)</td>
<td>111</td>
<td>351</td>
</tr>
<tr>
<td><strong>Evaluation Samples</strong></td>
<td></td>
<td></td>
</tr>
<tr>
<td>connector terms</td>
<td>111</td>
<td>354</td>
</tr>
<tr>
<td>PC coating selector</td>
<td>110</td>
<td>344</td>
</tr>
<tr>
<td>resins folder</td>
<td>110</td>
<td>343</td>
</tr>
<tr>
<td>resistors, chip</td>
<td>112</td>
<td>362</td>
</tr>
<tr>
<td>transformer guide</td>
<td>112</td>
<td>364</td>
</tr>
</tbody>
</table>

**Design Aids**
- design memoirs (DA) 110 342
- PC coating selector (NL) 110 344
- resins folder (NL) 110 343

**Production**
- blasting system 108 293
- fluidic devices (NL) 112 365
- pickup, vacuum 108 294
- soldering systems 108 292
- standards guide (AN) 111 351
- wire stripper 108 291

**New Literature**
- aerosols, electronics 112 362
- capacitors, military 112 359
- connectors 114 368
- display terms (NL) 112 360
- equipment rental 112 361
- fluidic devices (NL) 112 365
- Heathkit catalog (NL) 112 356
- microwave products (NL) 112 358
- modules, Ku-band 111 355
- standards guide (AN) 111 351

**Evaluation Samples**
- connector terms 111 354
- cyrogenics 111 352
- magnets, permanent 111 353
- multipliers, Ku-band 111 355
- standards guide (AN) 111 351

**Design Aids**
- design memoirs (DA) 110 342
- PC coating selector (NL) 110 344
- resins folder (NL) 110 343
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Match these RCA Op Amps to your design needs

<table>
<thead>
<tr>
<th>Low-noise Types and Economy Types</th>
<th>High-Gain, High-Power Types</th>
</tr>
</thead>
<tbody>
<tr>
<td>± 6-V types</td>
<td>± 12-V types</td>
</tr>
<tr>
<td>Flat Pack (14-lead)</td>
<td>CA3008 $3.90</td>
</tr>
<tr>
<td>TS (12-lead)</td>
<td>CA3010 $1.58</td>
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<td>DIC (14-lead)</td>
<td>CA3037 $1.90</td>
</tr>
<tr>
<td>DIP (14-lead)</td>
<td>CA3029 $ .98</td>
</tr>
<tr>
<td>Open Loop Voltage Gain</td>
<td></td>
</tr>
<tr>
<td>CMRR</td>
<td>60 dB typ.</td>
</tr>
<tr>
<td>Input Bias Current</td>
<td>12 μA max.</td>
</tr>
<tr>
<td>Input Impedance</td>
<td>10 kΩ min.</td>
</tr>
<tr>
<td>Noise Figure @ 1kHz</td>
<td>12 dB max.</td>
</tr>
</tbody>
</table>

(all prices shown at 1000-unit level)

INFORMATION RETRIEVAL NUMBER 214